

# Recent Advances in Experiments and Modeling of Grid-forming Systems

Brian Johnson

Assistant Professor

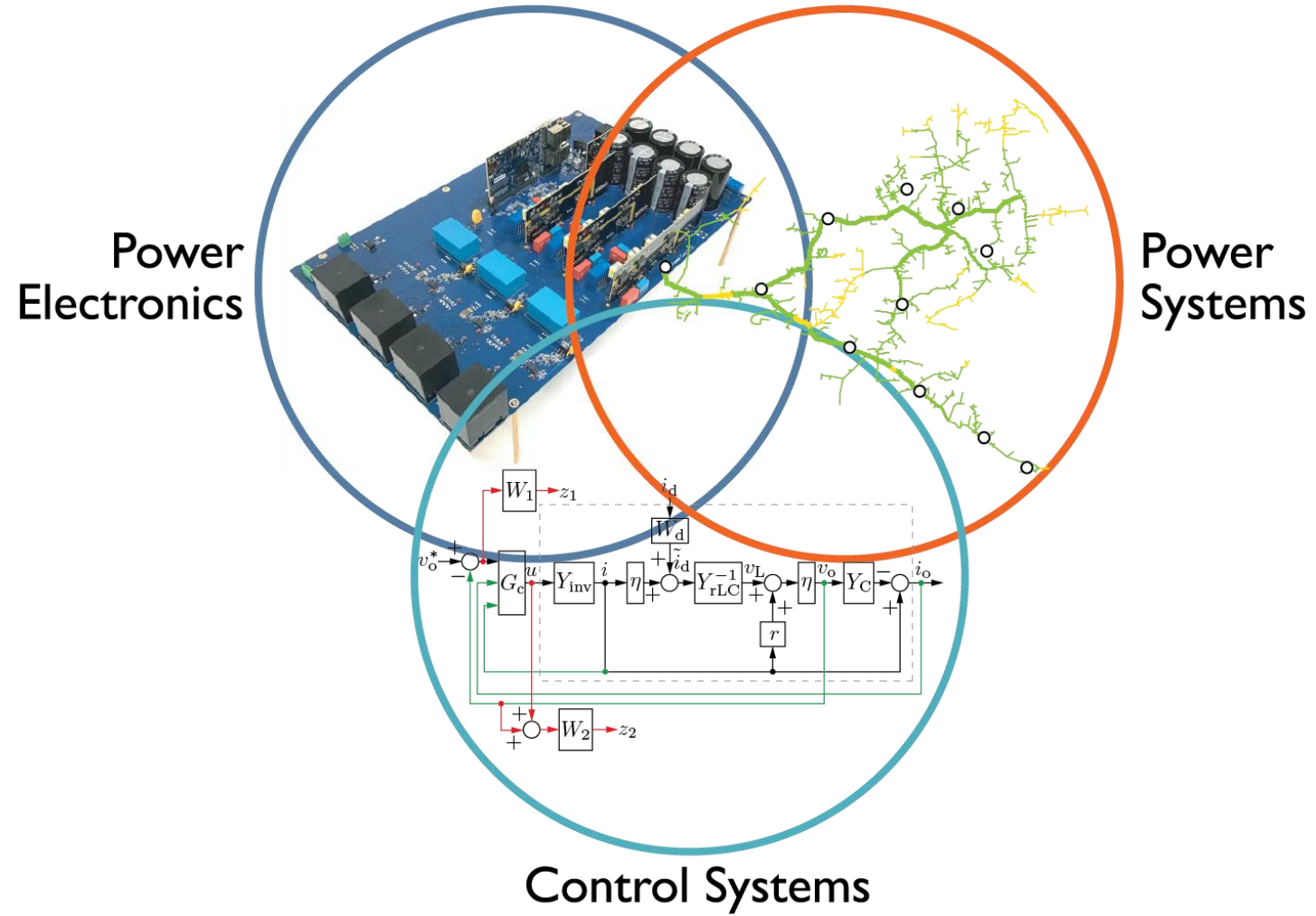
University of Texas at Austin

May 25<sup>th</sup>, 2023



The University of Texas at Austin  
**Chandra Department of Electrical  
and Computer Engineering**  
*Cockrell School of Engineering*

# My Lab Activities



# Acknowledgements

Contributions from my postdoc and 6 grad students



Minghui



Kamakshi



Rahul



Weiqian



Soham



Pranav



Debjyoti

and generous support from DOE & NSF for my research on:

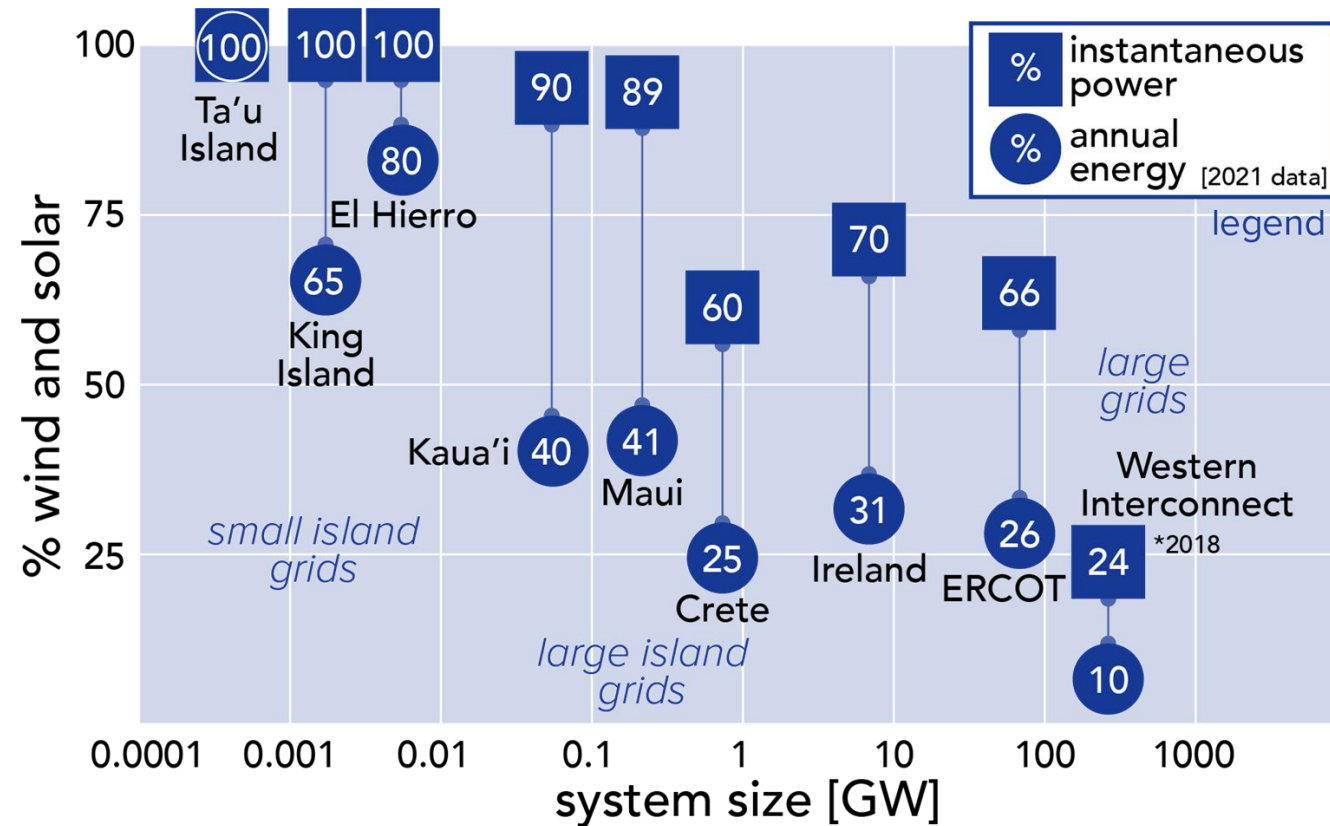
- Grid-forming systems
- Power electronics
- UNIFI Consortium



U.S. DEPARTMENT OF  
**ENERGY**

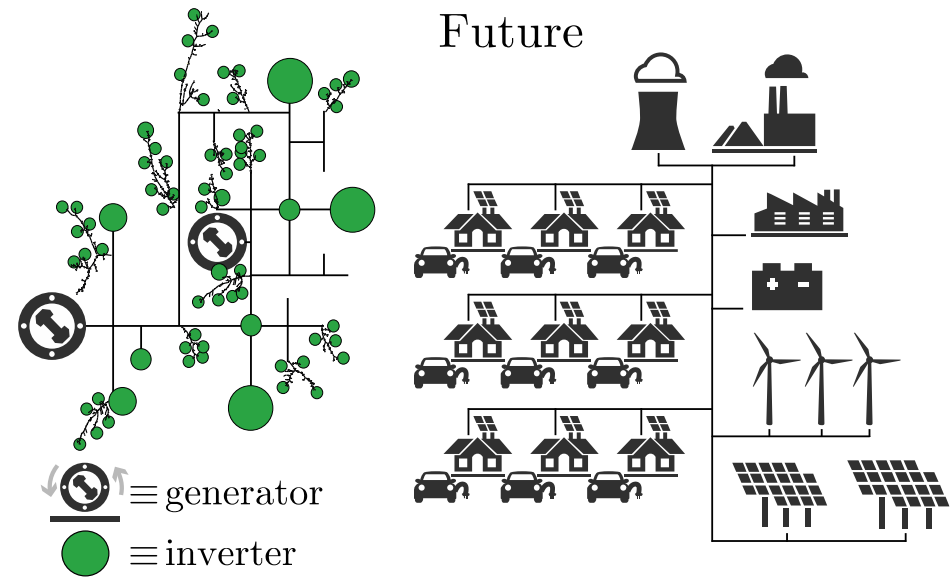
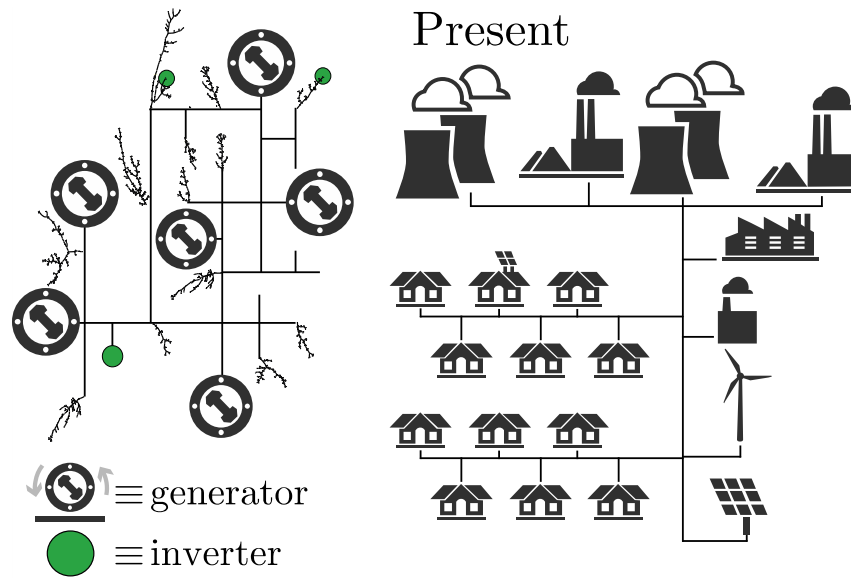
Energy Efficiency &  
Renewable Energy

# Renewable Utilization for Various Grid Sizes

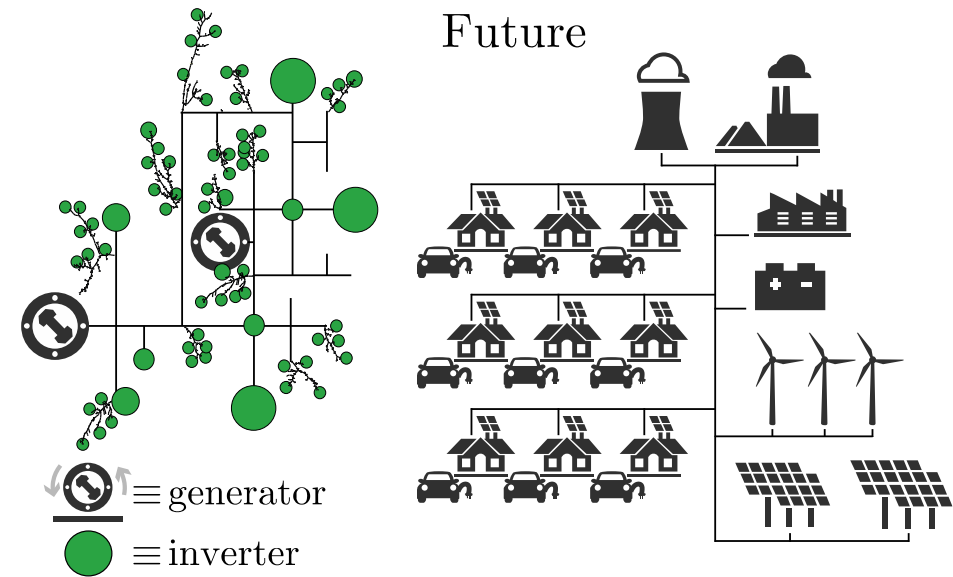
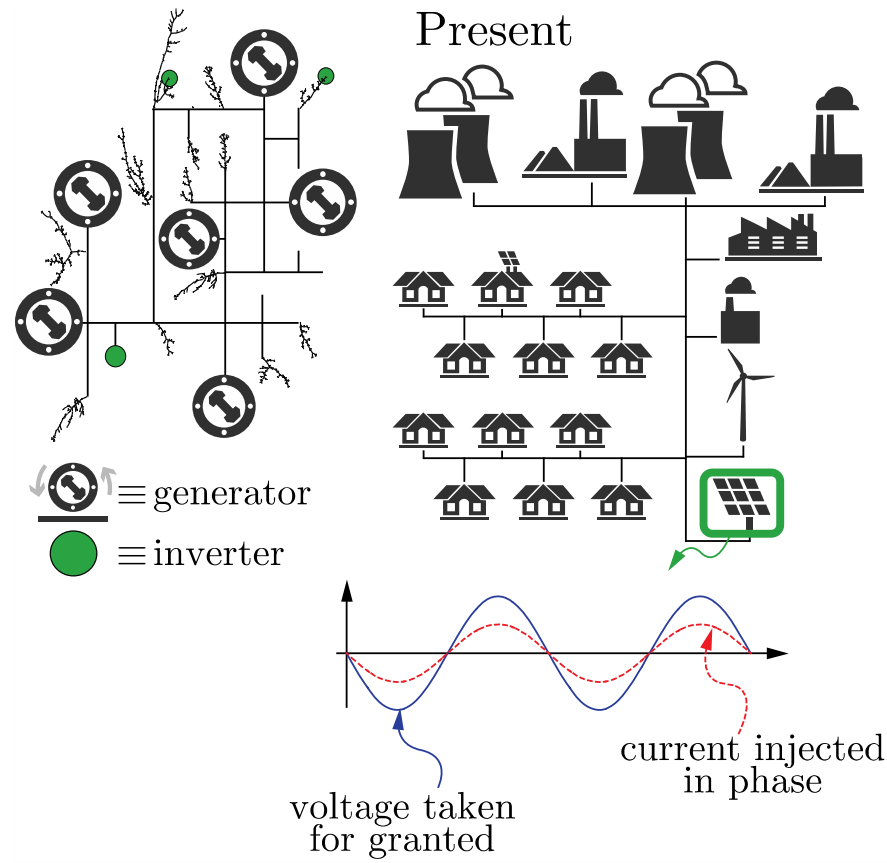


- Engineering challenges grow with system size and complexity
- Need scalable, robust, and resilient methods for system operation

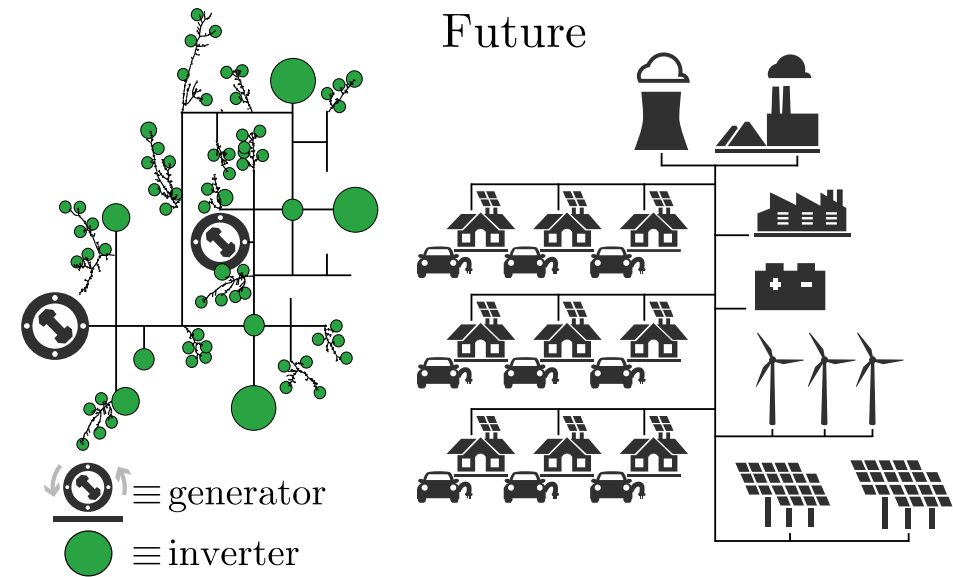
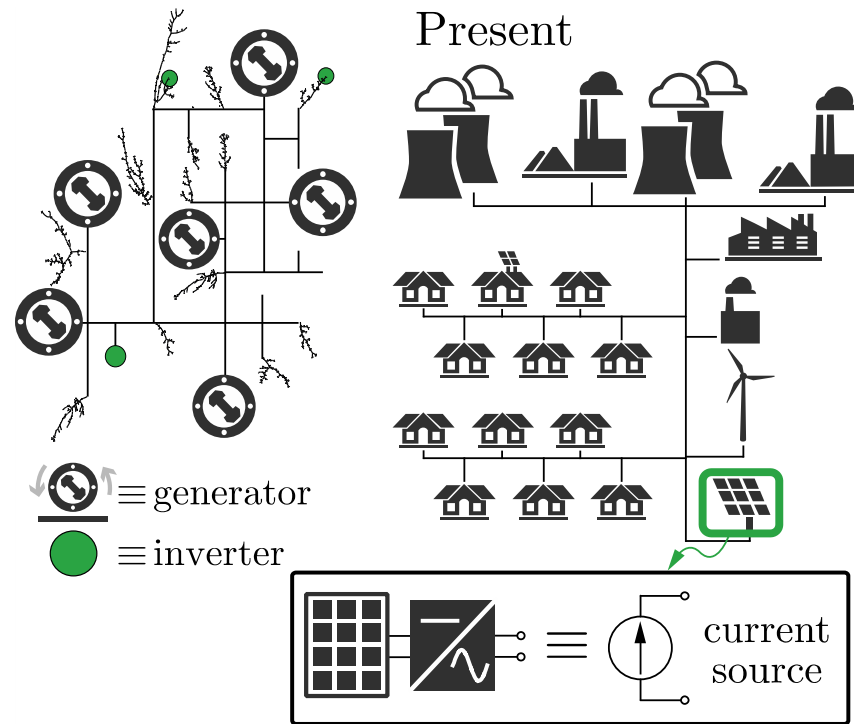
# A Vision for the Future Grid



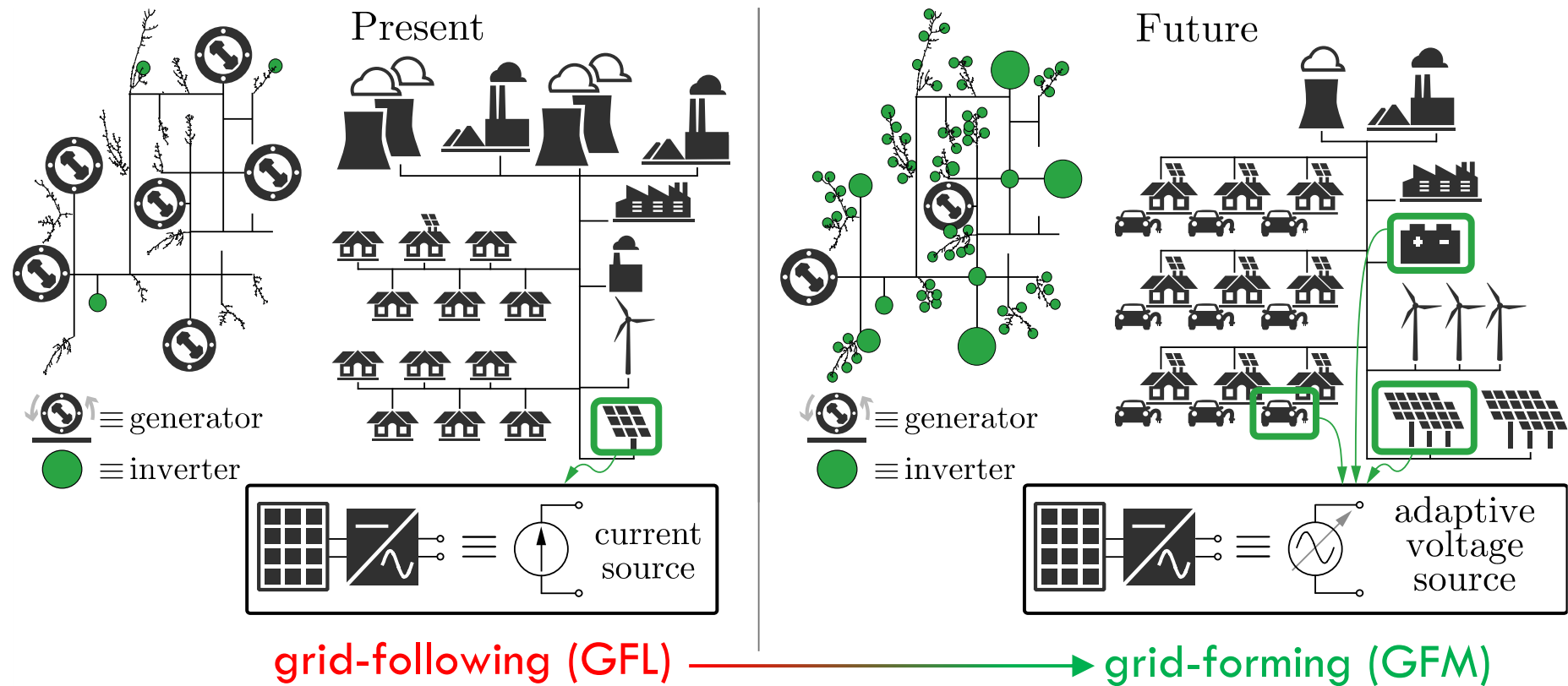
# A Vision for the Future Grid



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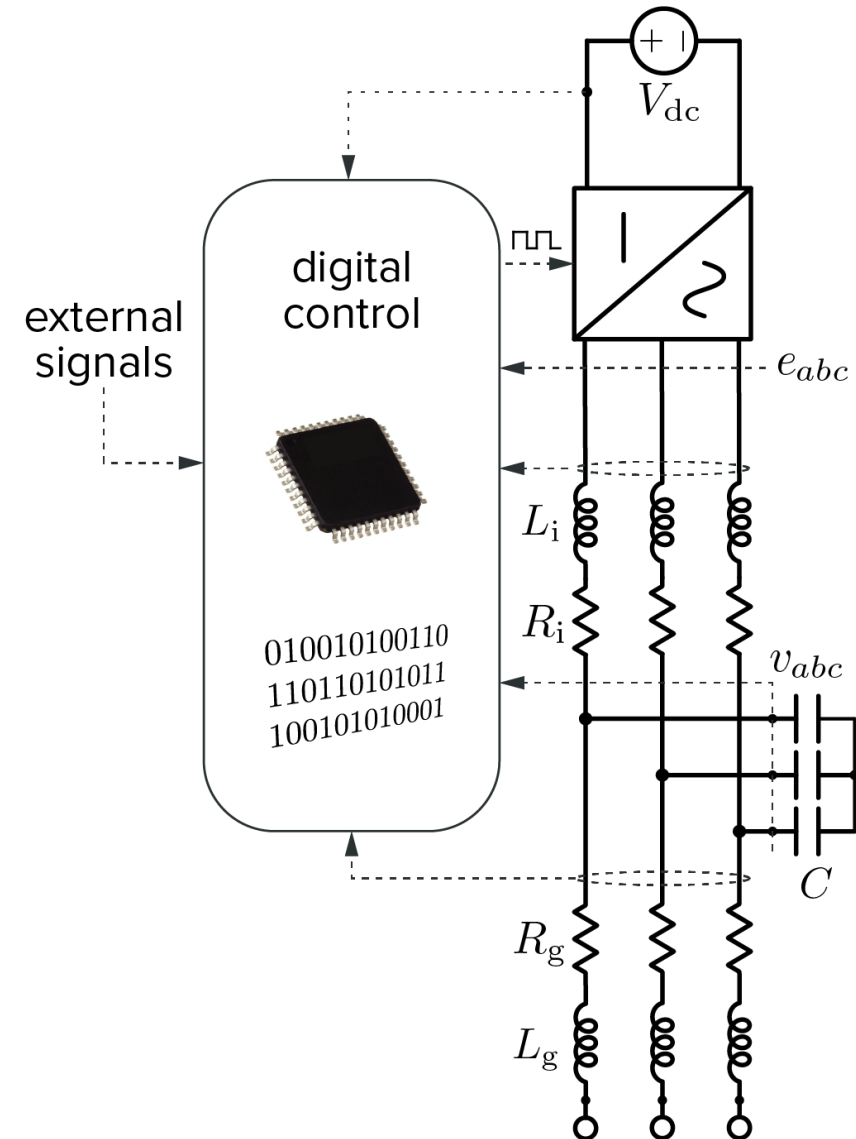
Achieving 2 goals simultaneously:

- Break down barriers that limit adoption of renewable energy
- Realize a bottom-up system that works resiliently at any scale

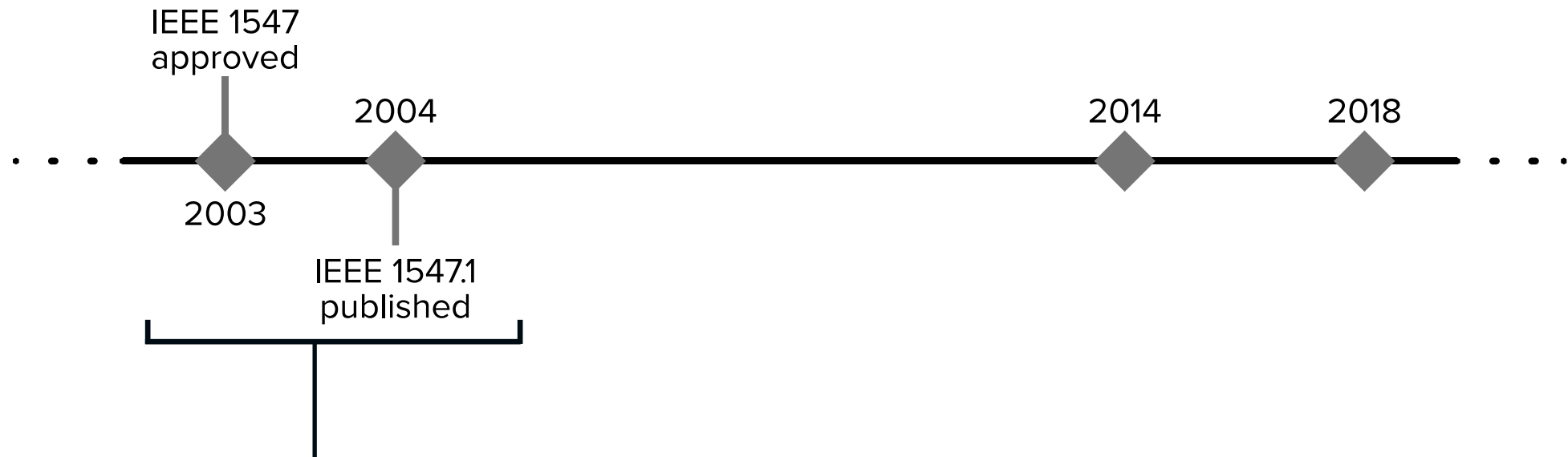


# Desired Characteristics of a GFM Inverter

- Can operate in standalone or with many other GFM
- Proportional power sharing among units
- Works when connected to a stiff or weak grids
- Automatic synchronization
- Exogenous signals can adjust power delivery
- Has current limiter to prevent overcurrents
- Integrates with dc-side controls



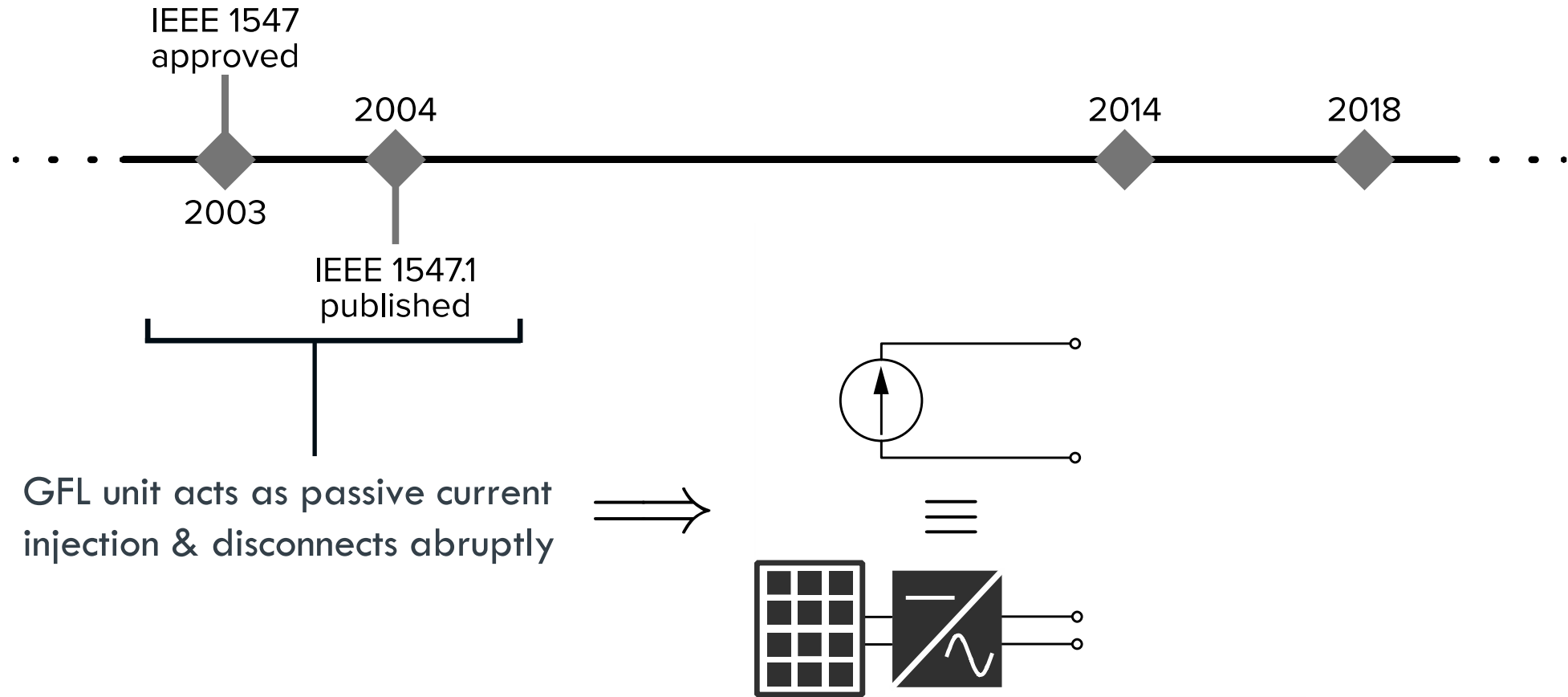
# An Evolving Regulatory Landscape



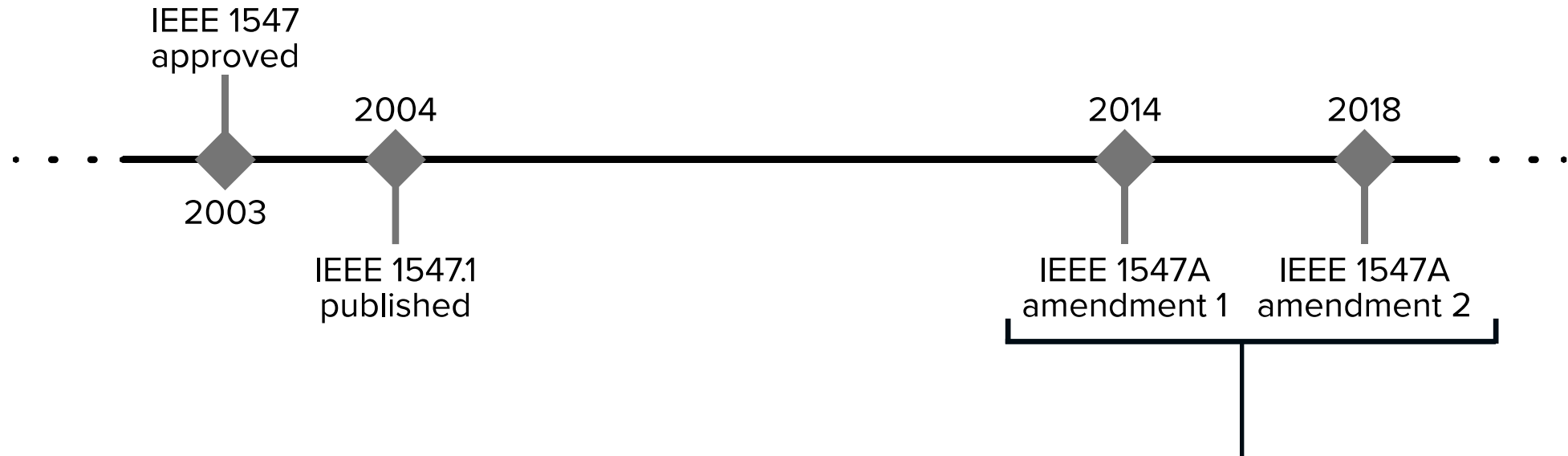
The inverter shall:

- NOT regulate voltage
- TRIP under abnormal voltage/frequency

# An Evolving Regulatory Landscape



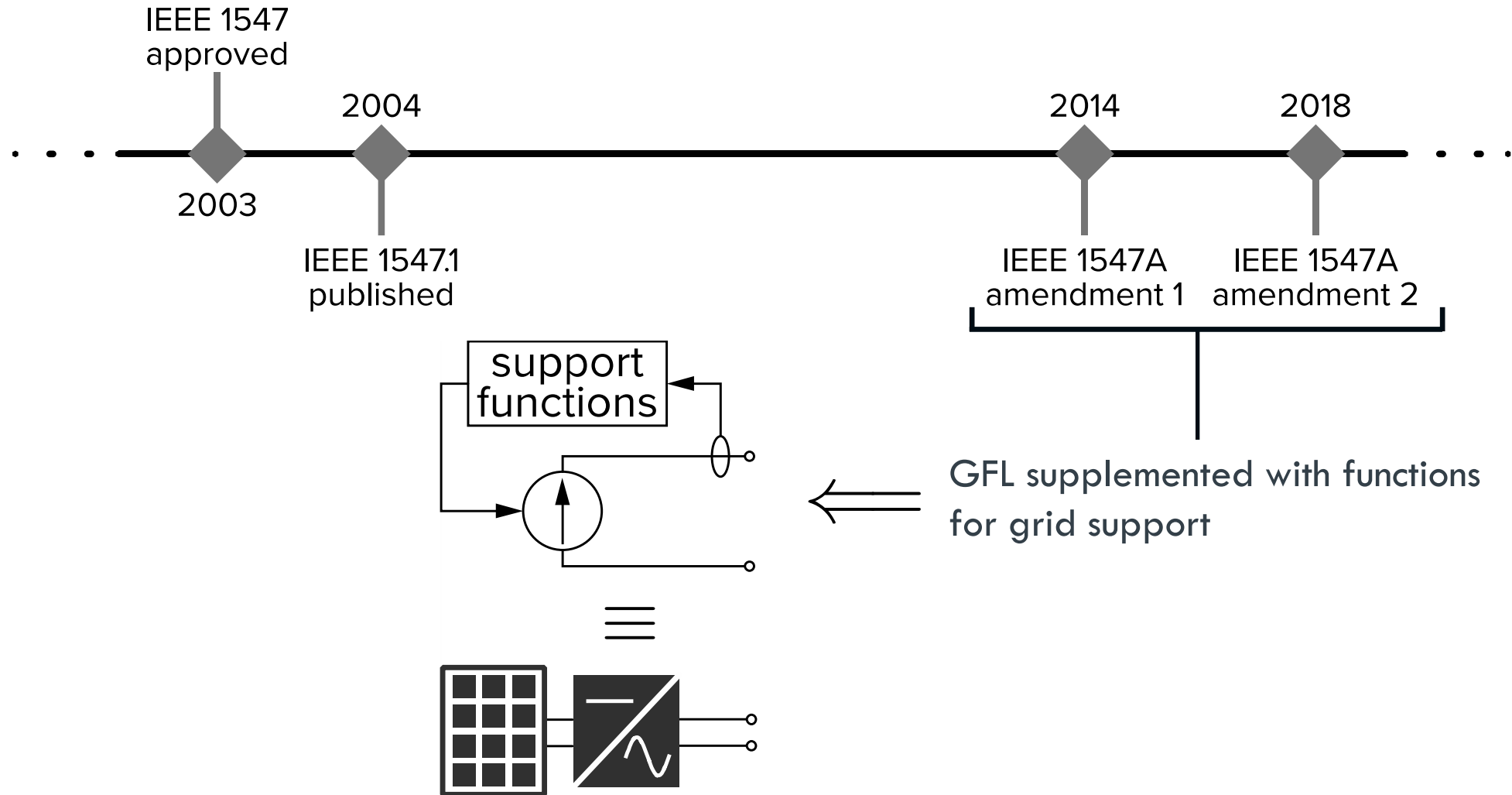
# An Evolving Regulatory Landscape



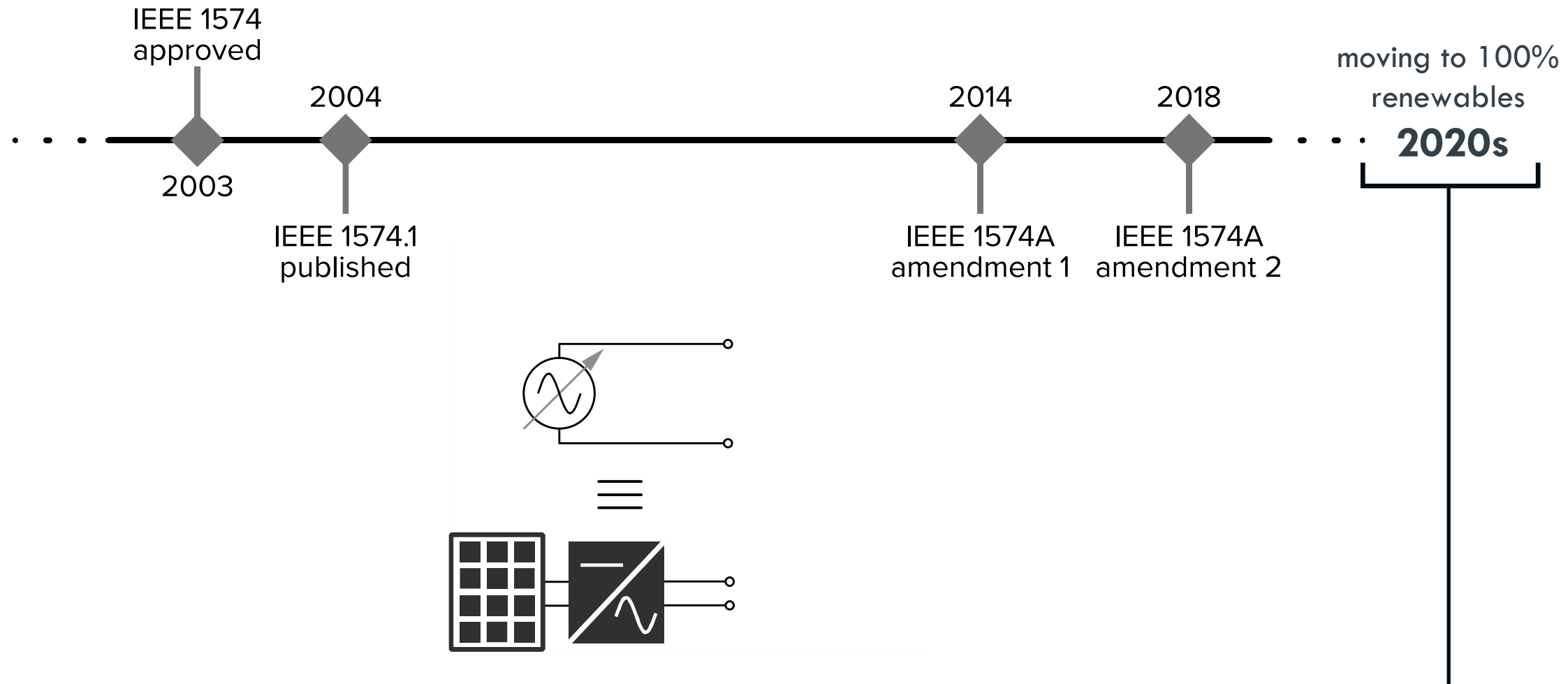
The inverter shall:

- Assist voltage regulation
- Ride-through abnormal voltage/frequency

# An Evolving Regulatory Landscape



# An Evolving Regulatory Landscape



Full-fledged grid-forming (GFM) functionality needed to realize ambitious renewable targets<sup>[1]</sup>

[1] List of territories, states, cities, communities committed to 100% renewable energy: [www.sierraclub.org/ready-for-100](http://www.sierraclub.org/ready-for-100)

# Remaking the Electric Power Industry

The organization that will bring GFM technologies onto grids:

- Duration: 2022-2027
- Co-Leads: UT, NREL, EPRI
- Total Funds: \$34.9M
- Federal Funds: \$25M
- Membership:
  - 12 universities
  - 4 national labs
  - 25 industry members

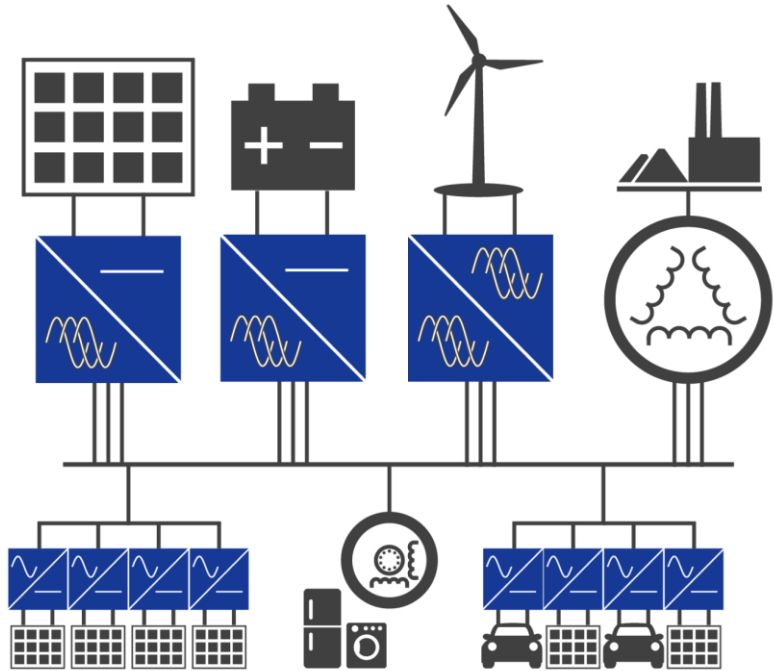


Funded by:



**SOLAR ENERGY  
TECHNOLOGIES OFFICE**  
U.S. Department Of Energy

# Unifying Technologies Across All Scales



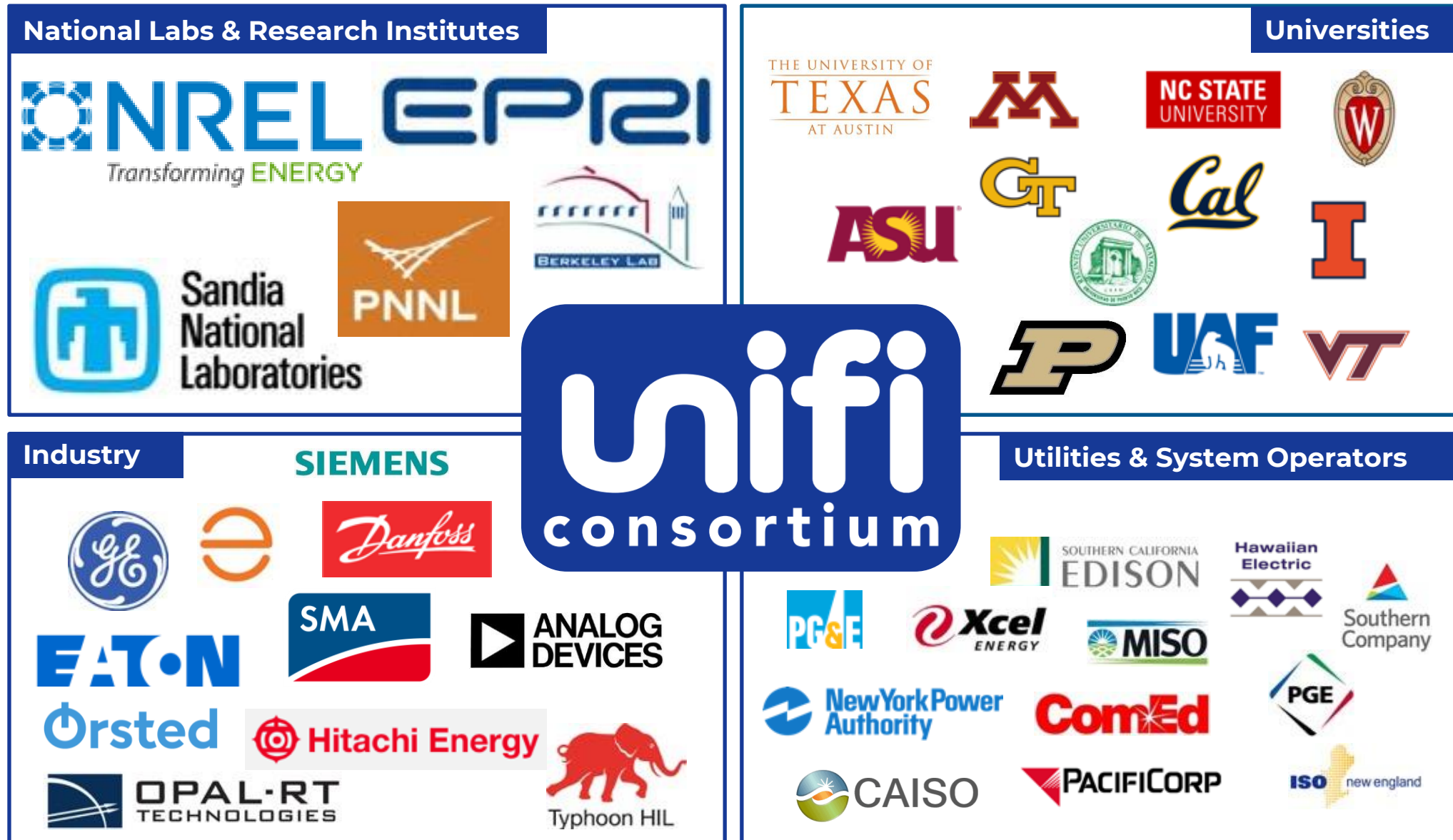


- New & Old (GFM with GFL+Machines)
- Local & Global (controls)
- Slow & Fast (timescales of operation)
- Big & Small (inverters to aggregations)
- Solar, Wind, & Storage (technologies)





# A Team with Crosscutting Perspectives Across the Electric Power Industry



# A Team with Crosscutting Perspectives Across the Electric Power Industry

These + many others will need to work towards consensus



SIEMENS

EAT•N

*Danfoss*


 Hitachi Energy

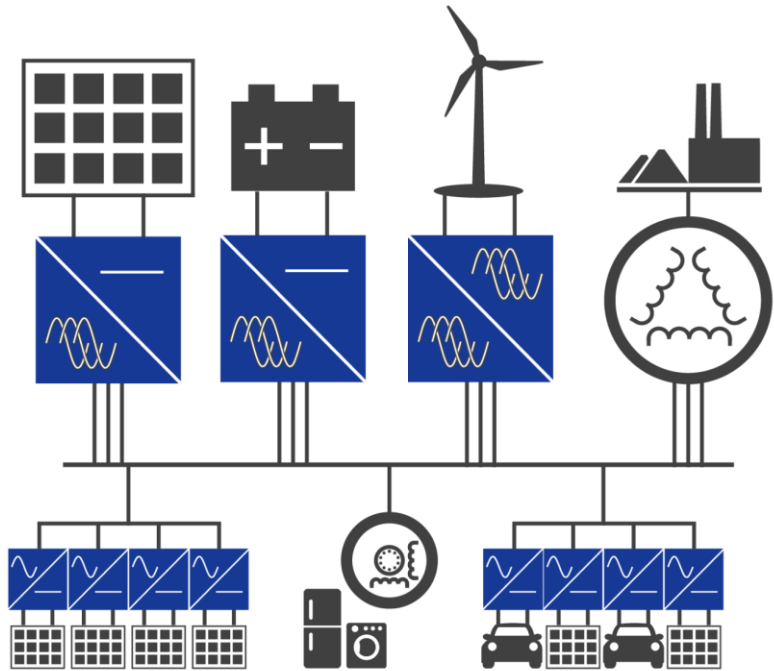
SMA

Ørsted



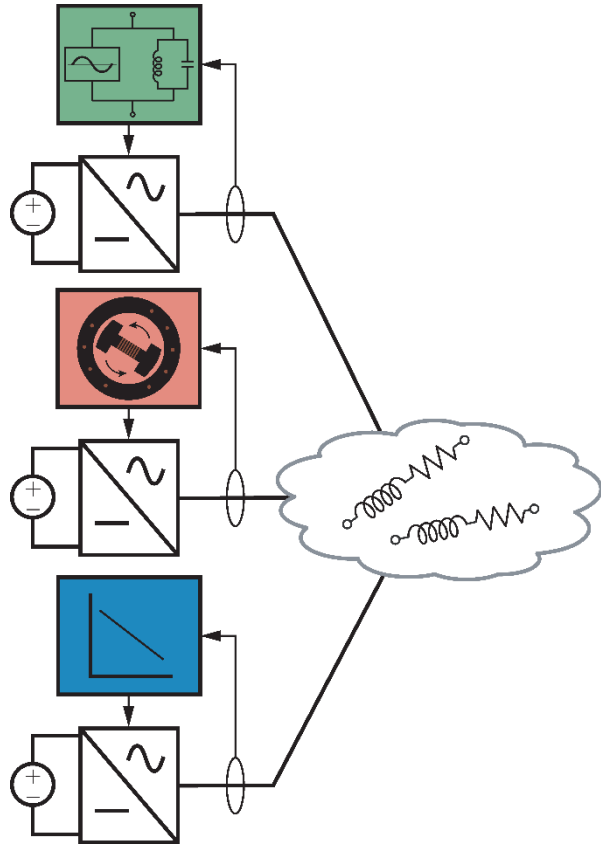
# Universal Interoperable Operation is Essential

Seeking inspiration from  **Bluetooth**<sup>®</sup>  
& cultivating an ecosystem of innovation

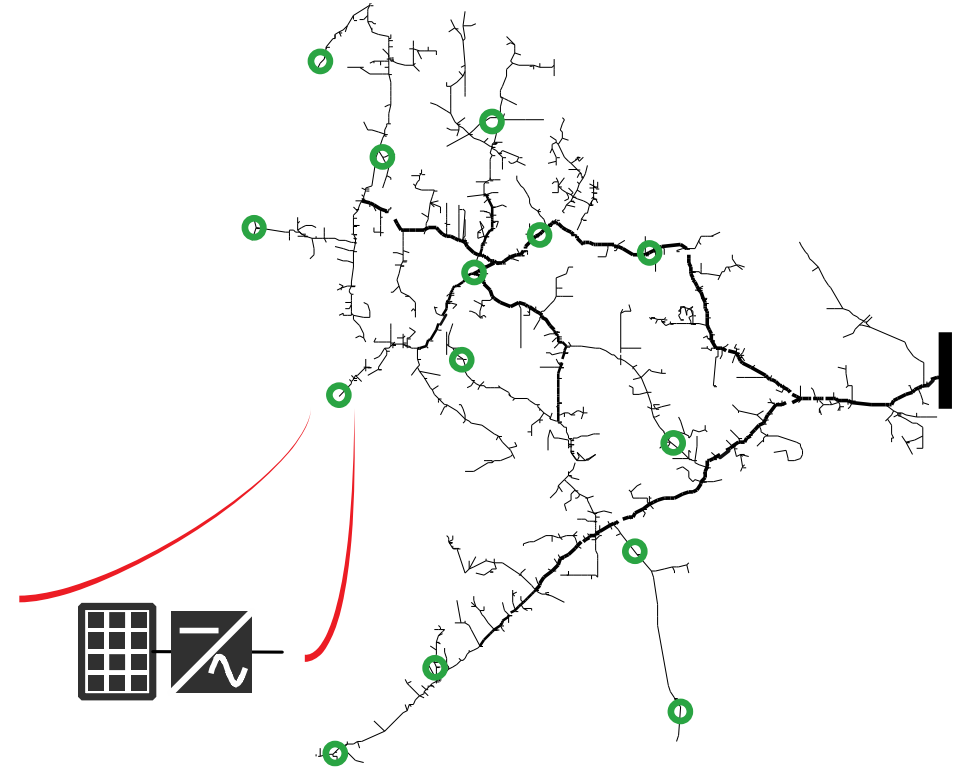


Interoperable GFM controls will

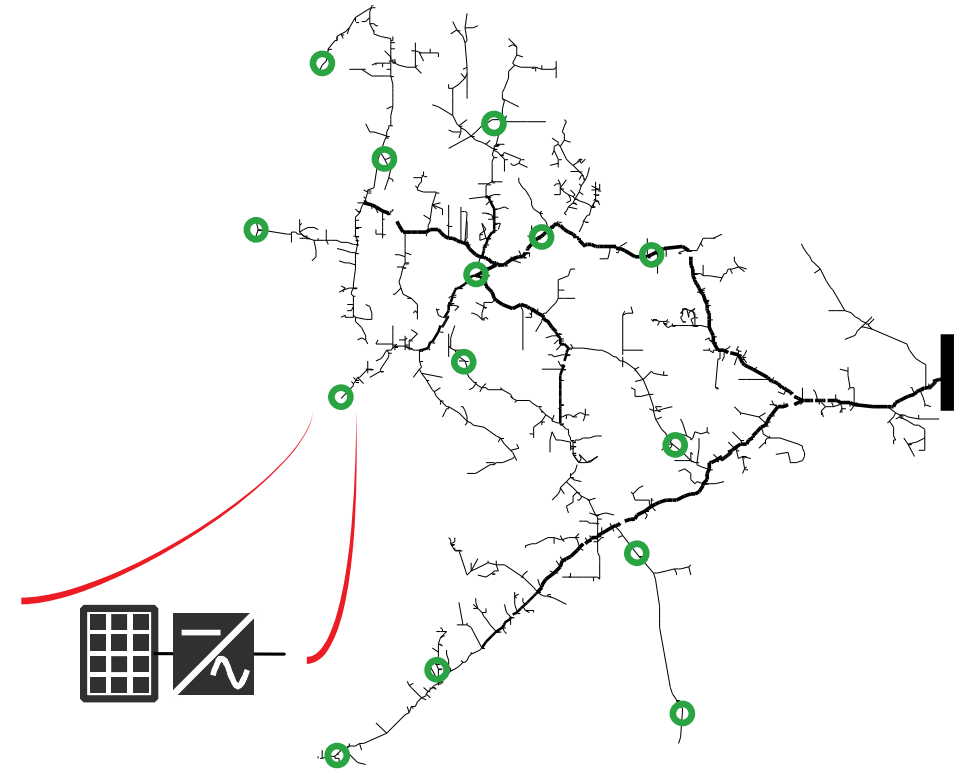
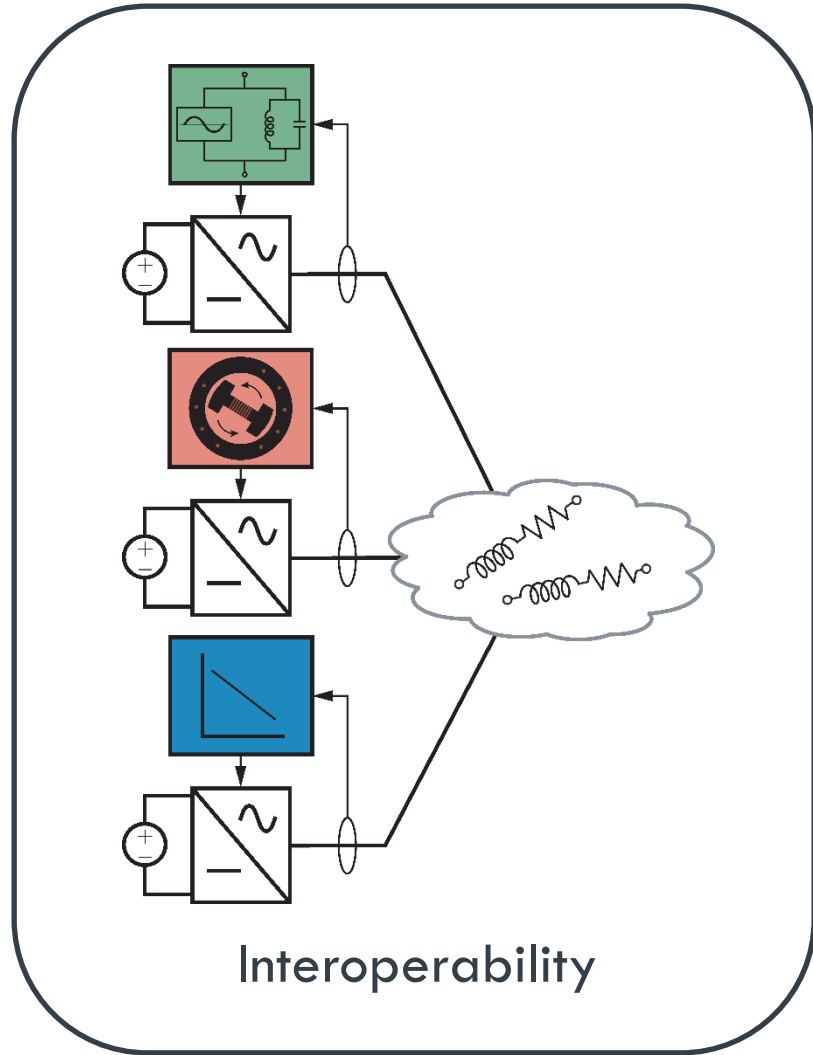
- Enable all manufacturers to innovate
- Coexist with proprietary functions
- Streamline first-principles-based standards
- Be black-box testable for certification



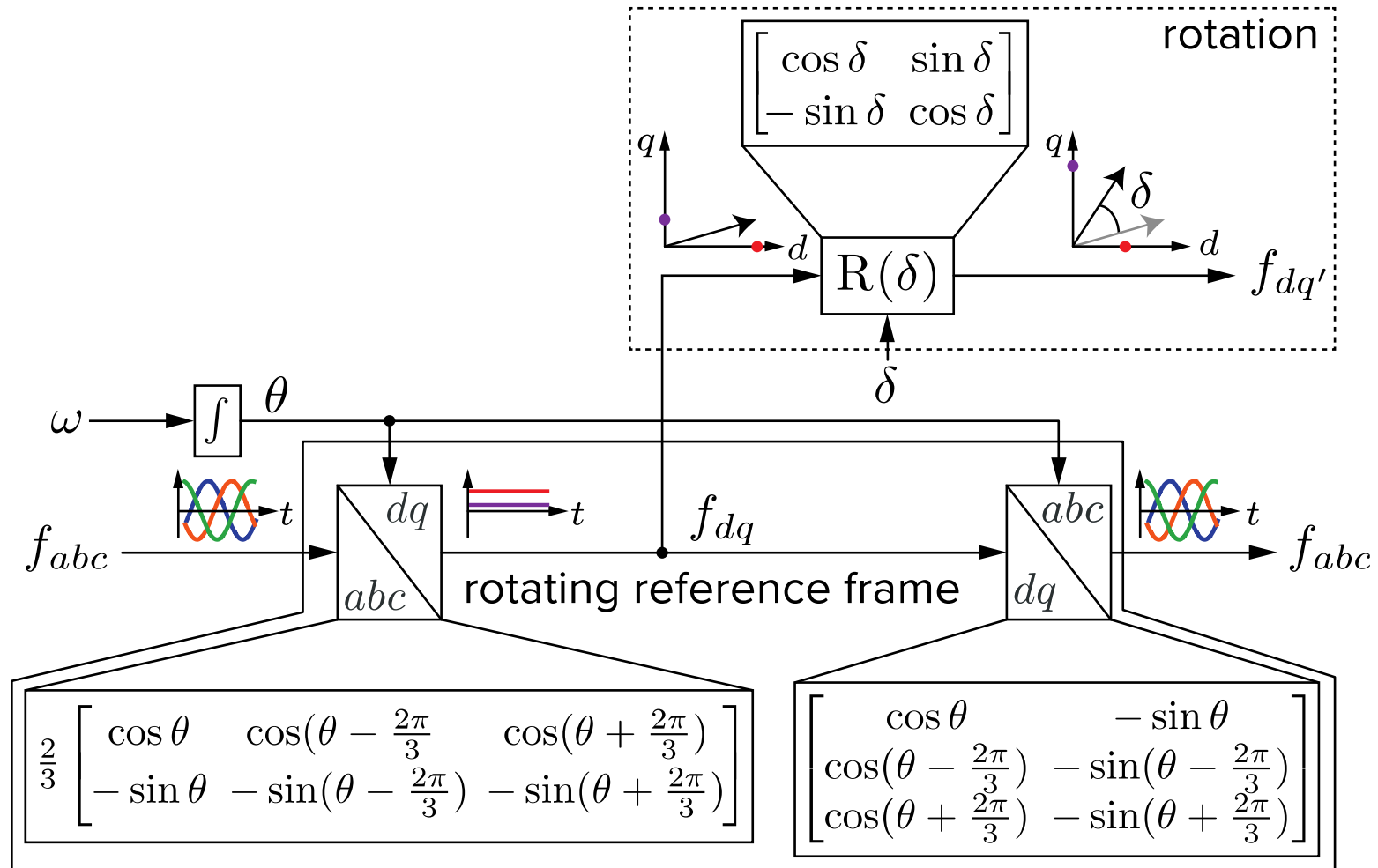
Interoperability



Self-organizing GFM Networks

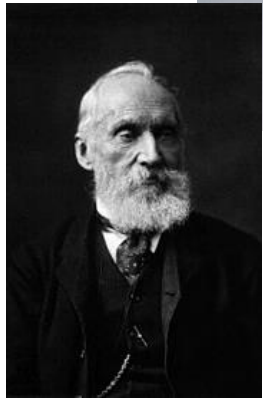


# Reference Frame Basics for Three-phase Systems

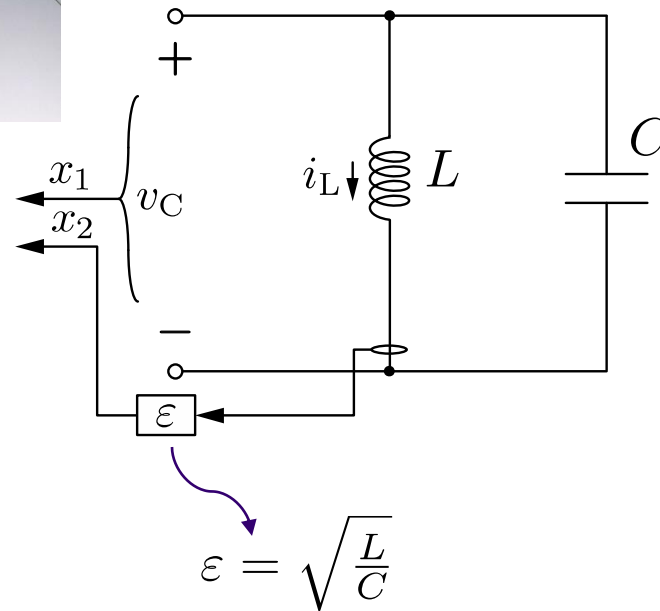


# We Begin with a Humble Little Circuit

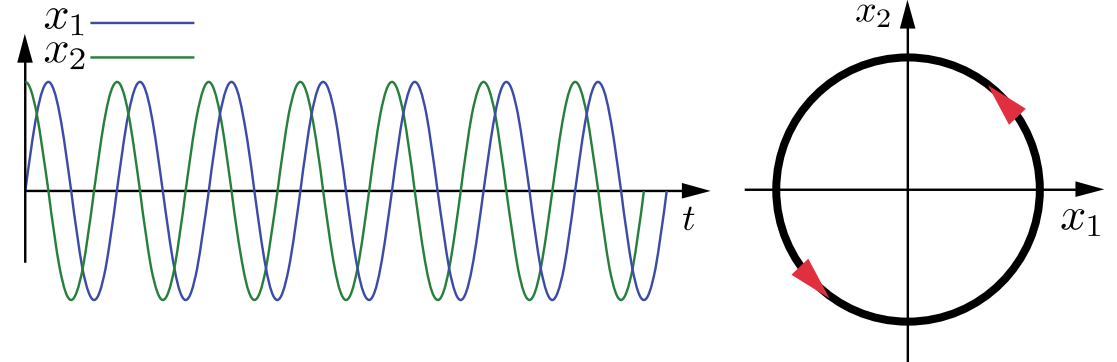
The simple harmonic oscillator lies at the heart of modern GFM control



William Thomson  
1824-1907



$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & -\omega_0 \\ \omega_0 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

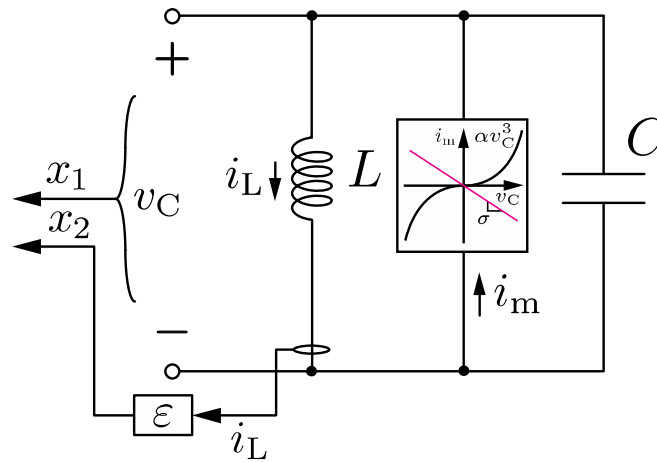


# Early Oscillator Models with Self-regulating Limit Cycles

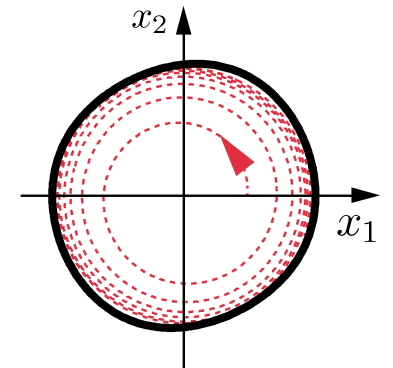
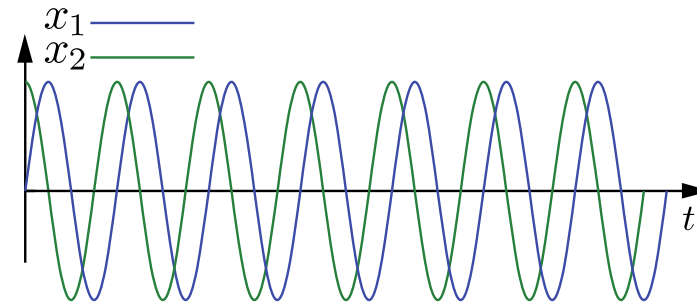
The van der Pol oscillator is a stepping-stone towards a power application



Balthasar van der Pol  
1889-1959



$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} \overbrace{\varepsilon \omega_0 (\sigma - \alpha x_1^2)}^{i_m/C} \\ \omega_0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$



[1] van der Pol, "On "relaxation-oscillations", The London, Edinburgh, and Dublin Philosophical Magazine and Journal of Science, 1926.

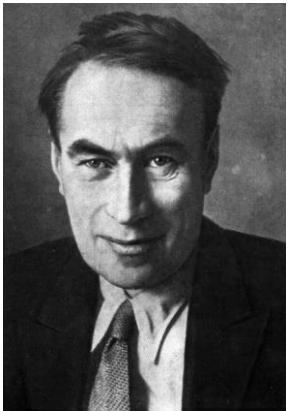
[2] B. Johnson, Dhople, Hamadeh, Krein, "Synchronization of parallel single-phase inverters using virtual oscillator control," TPEL, 2014.

[3] Awal, Yu, Tu, Lukic, Husain, "Hierarchical control for virtual oscillator based grid-connected and islanded microgrids," TPEL, 2020.

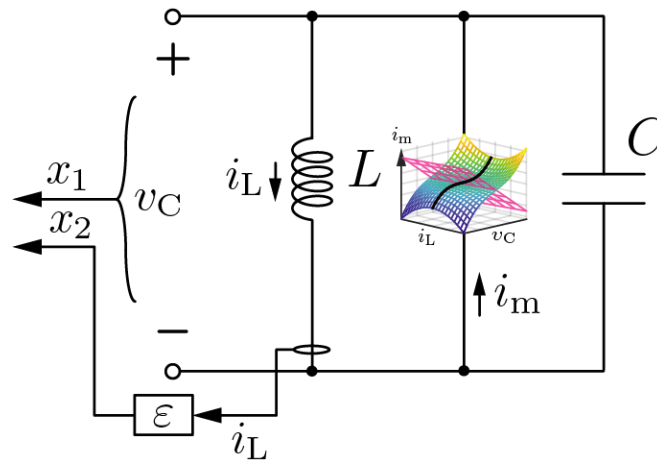


# Early Oscillator Models with Self-regulating Limit Cycles

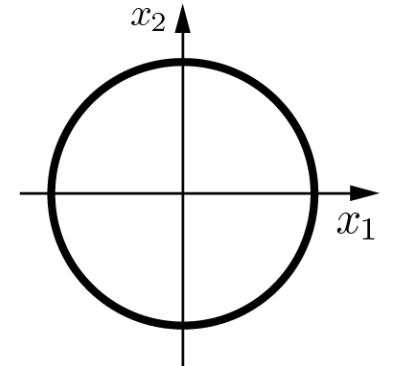
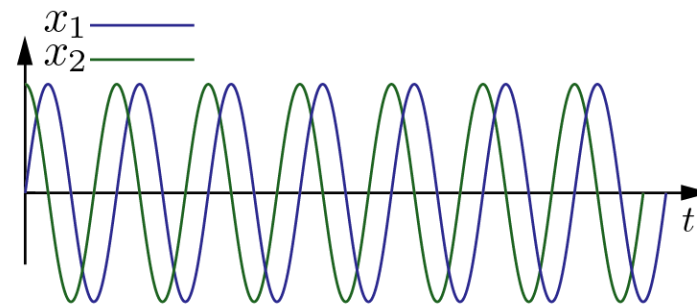
Andronov's oscillator gives harmonic-free waveforms useful for high power quality



Aleksandr Andronov  
1901-1952



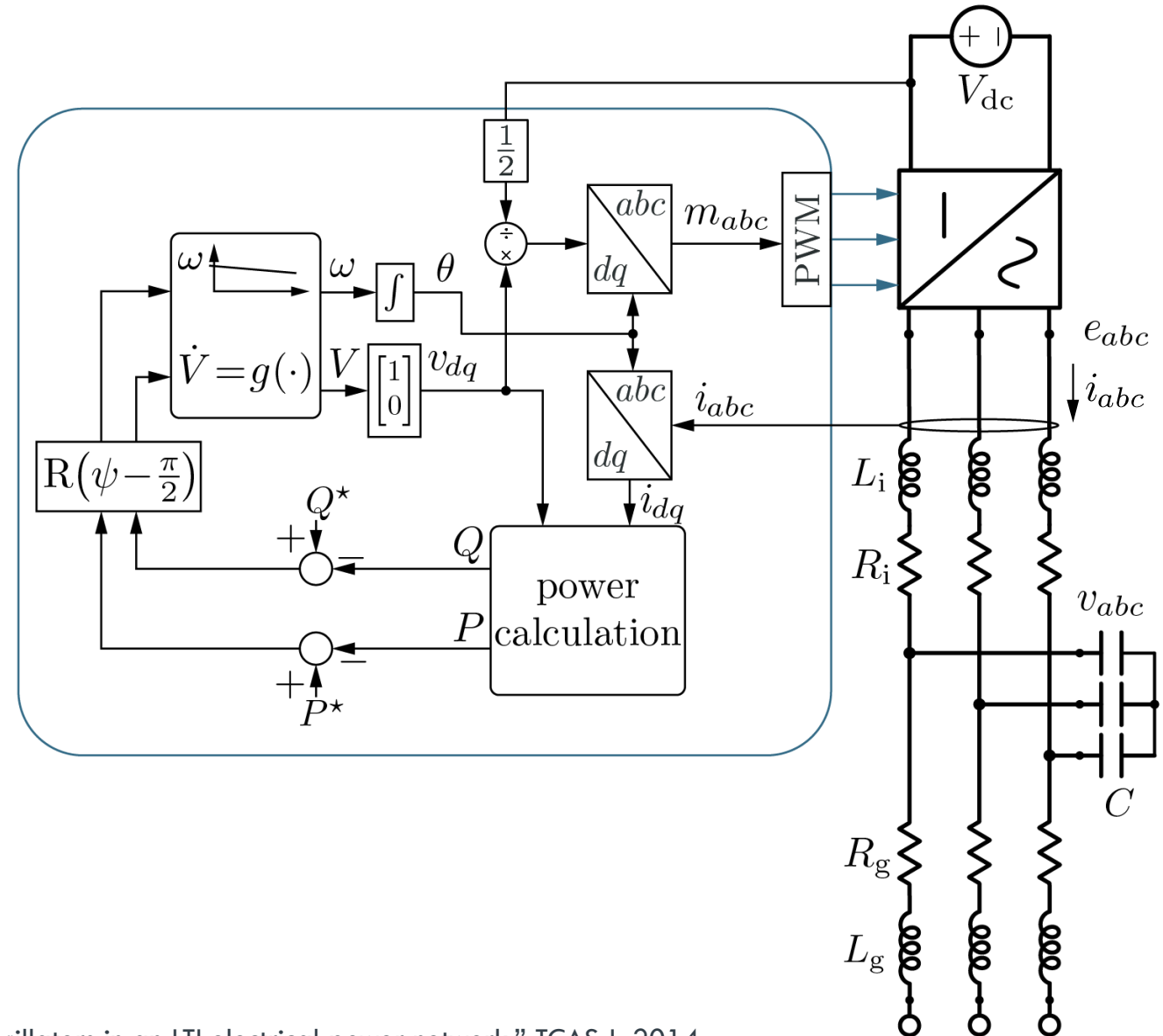
$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} \overbrace{\xi(2X_{\text{nom}}^2 - \|x\|^2)}^{i_m/C} \\ \omega_0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$



- [1] Andronov, Leontovich, Gordon, Maier, "Theory of Bifurcations of Dynamical Systems on a Plane," Israel Program Science Translation, 1971.
- [2] Columbino, Gross, Dorfler, "Global phase and voltage synchronization for power inverters: A decentralized consensus-inspired approach," CDC, 2017.
- [3] Lu, Dutta, Purba, Dhople, Johnson, "A grid-compatible virtual oscillator controller: Analysis and design" ECCE, 2019.

# Approach #1: Dispatchable Virtual Oscillator Control

- Transformed oscillator gives model here
- Newest GFM type in existence



[1] Johnson, Dhople, Hamadeh, Krein, "Synchronization of nonlinear oscillators in an LTI electrical power network," TCAS-I, 2014.

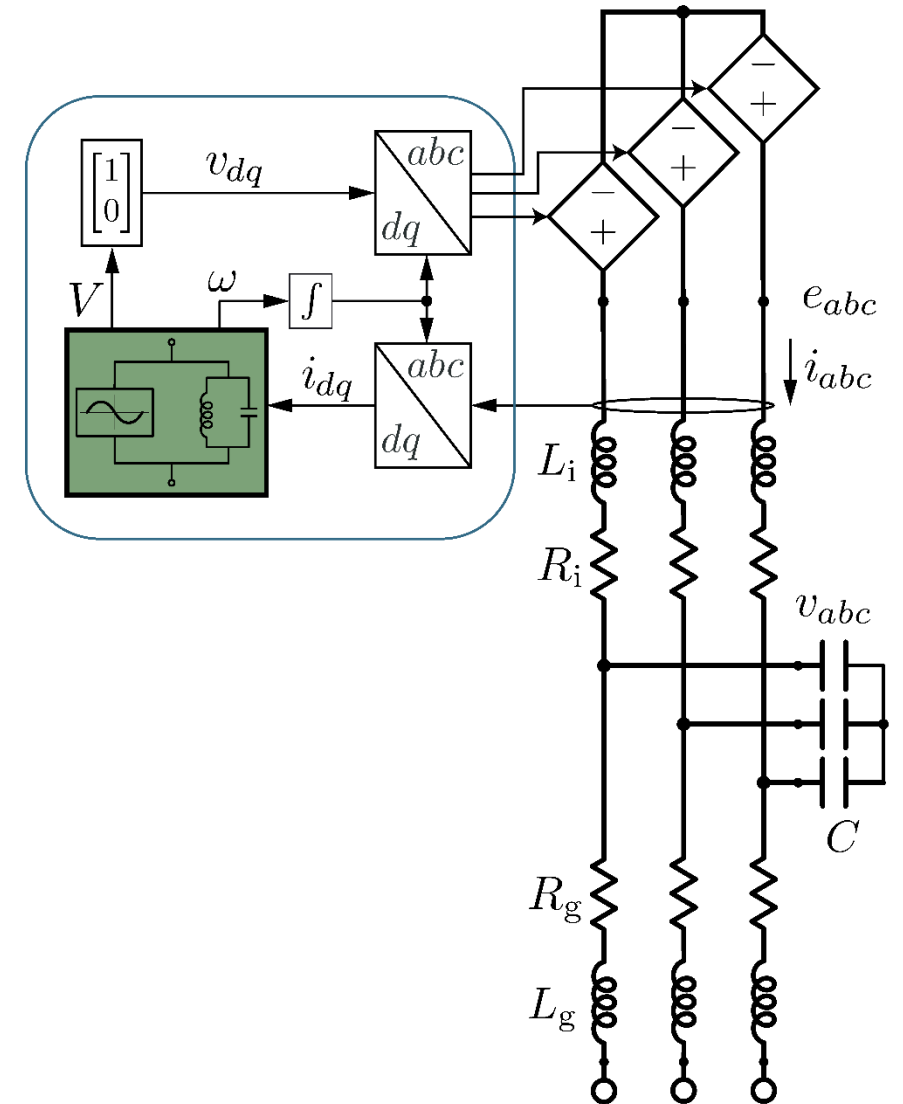
[2] Colombino, Groß, Dorfler, "Global phase and voltage synchronization for power inverters: A decentralized consensus-inspired approach," CDC, 2017.

# Approach #1: Dispatchable Virtual Oscillator Control

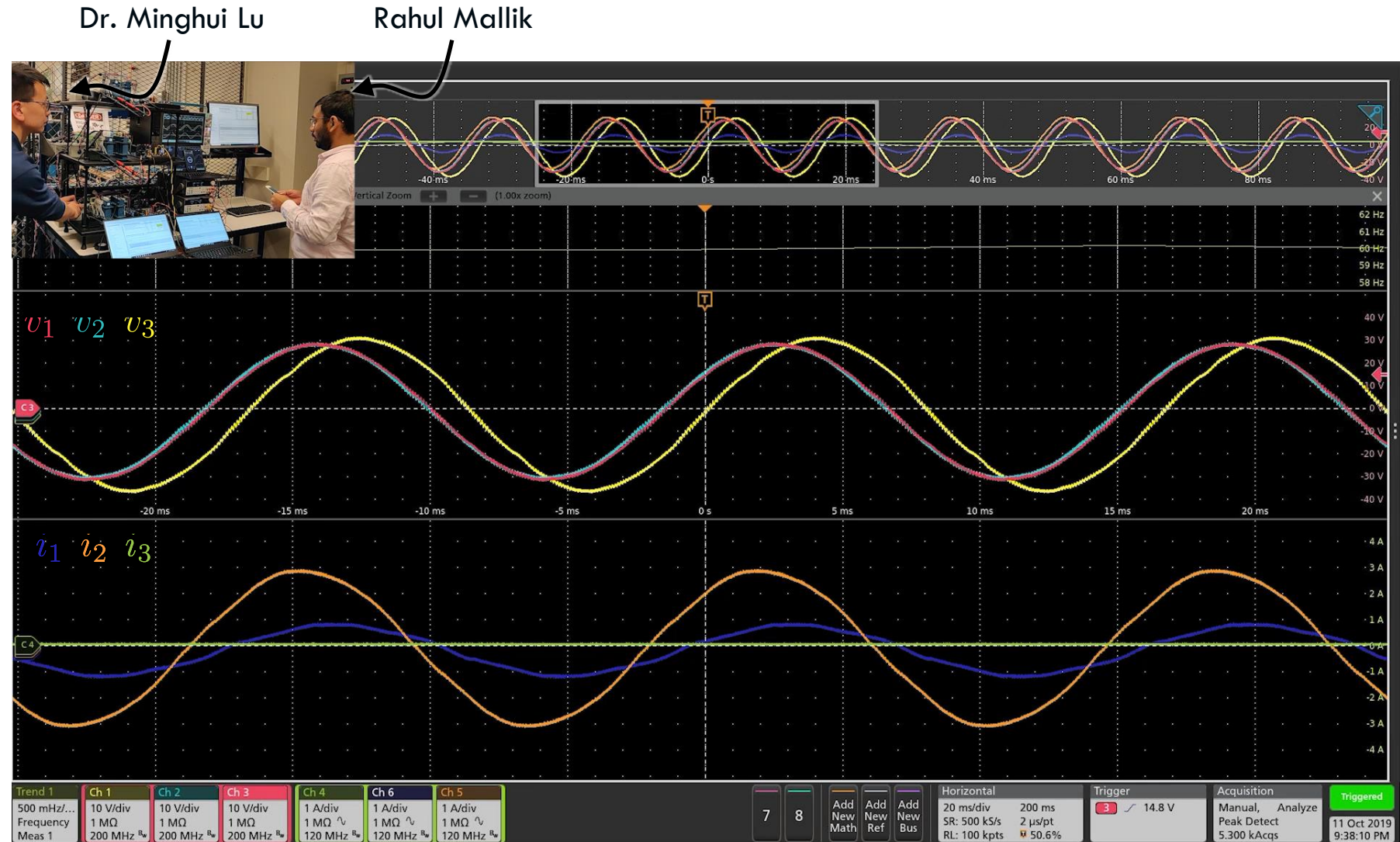
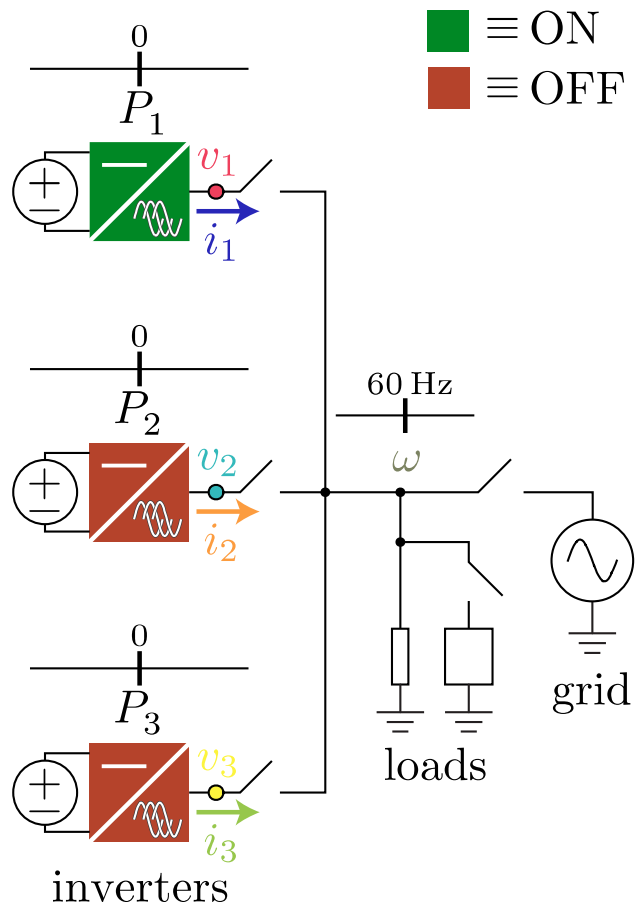
Control equations are given by

$$\omega = \omega_0 + \frac{\omega_0 \kappa_1}{V^2} \begin{bmatrix} 1 & 0 \end{bmatrix} \mathbf{R}(\psi - \frac{\pi}{2}) \begin{bmatrix} P^* - P \\ Q^* - Q \end{bmatrix},$$

$$\dot{V} = \omega_0 \kappa_2 V (V_0^2 - V^2) + \frac{\omega_0 \kappa_1}{V} \begin{bmatrix} 0 & 1 \end{bmatrix} \mathbf{R}(\psi - \frac{\pi}{2}) \begin{bmatrix} P^* - P \\ Q^* - Q \end{bmatrix}.$$

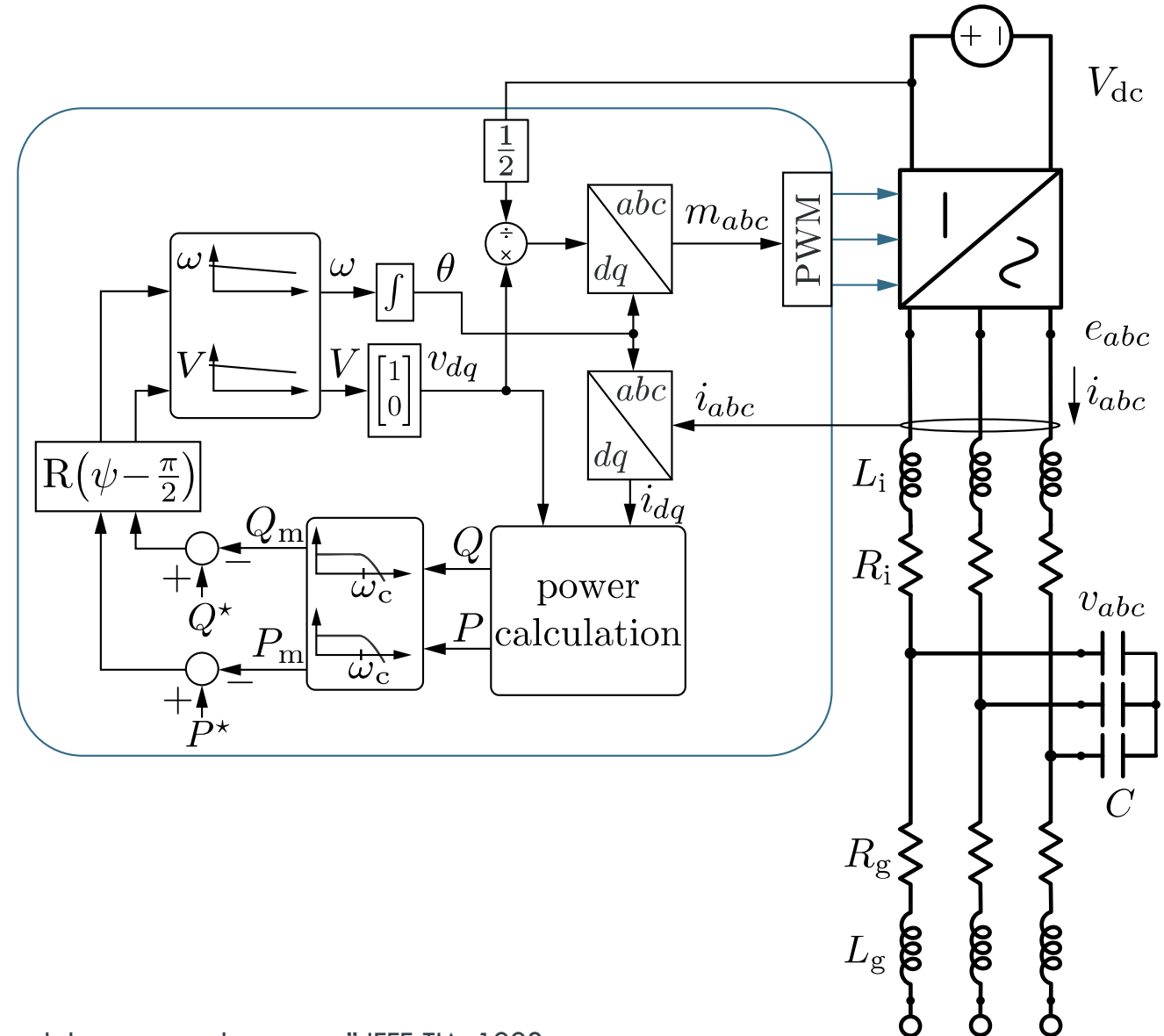


# Illustrating Versatile Performance on Hardware with dVOC Control



## Approach #2: Droop Control

- Oldest GFM method in existence
- Inspired by machine droop laws



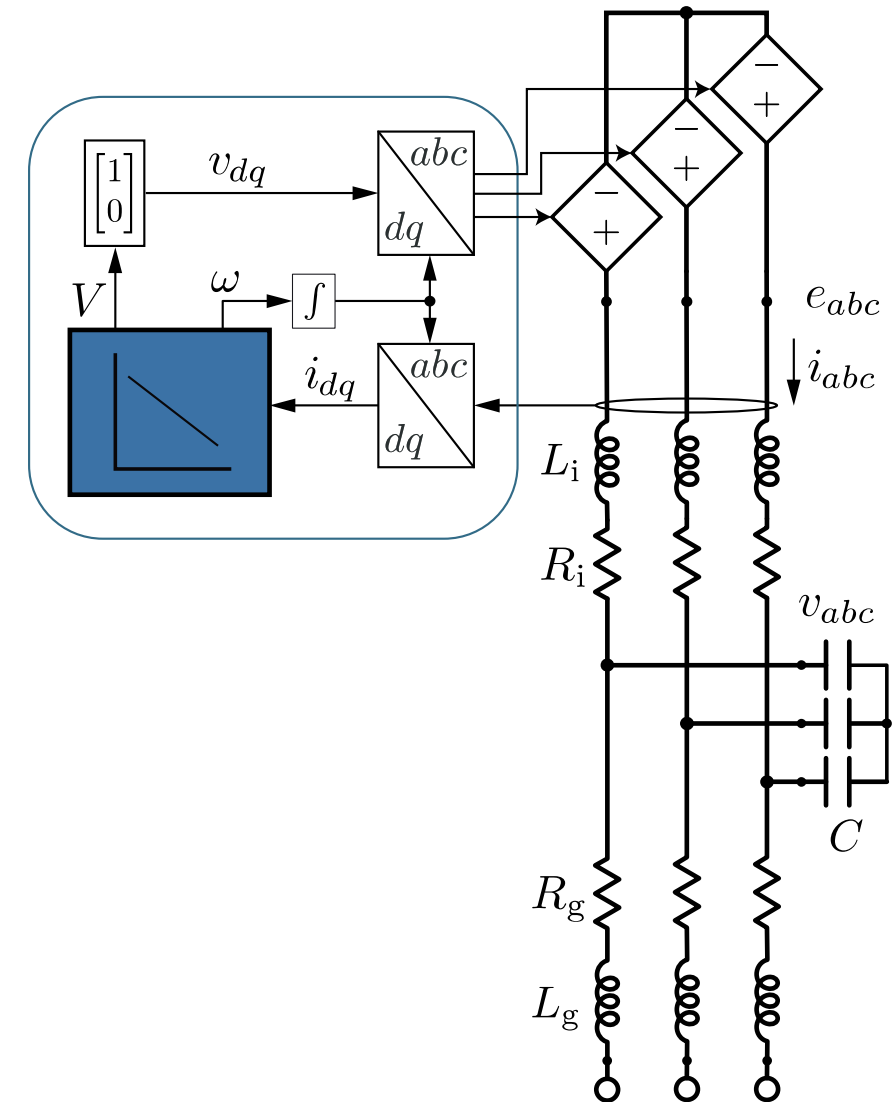
[1] M. Chandorkar, et. al, "Control of parallel connected inverters in standalone ac supply systems," IEEE TIA, 1993

[2] J. Guerrero, et. al, "Hierarchical control of droop-controlled dc and ac microgrids," IEEE TIE, 2011.

## Approach #2: Droop Control

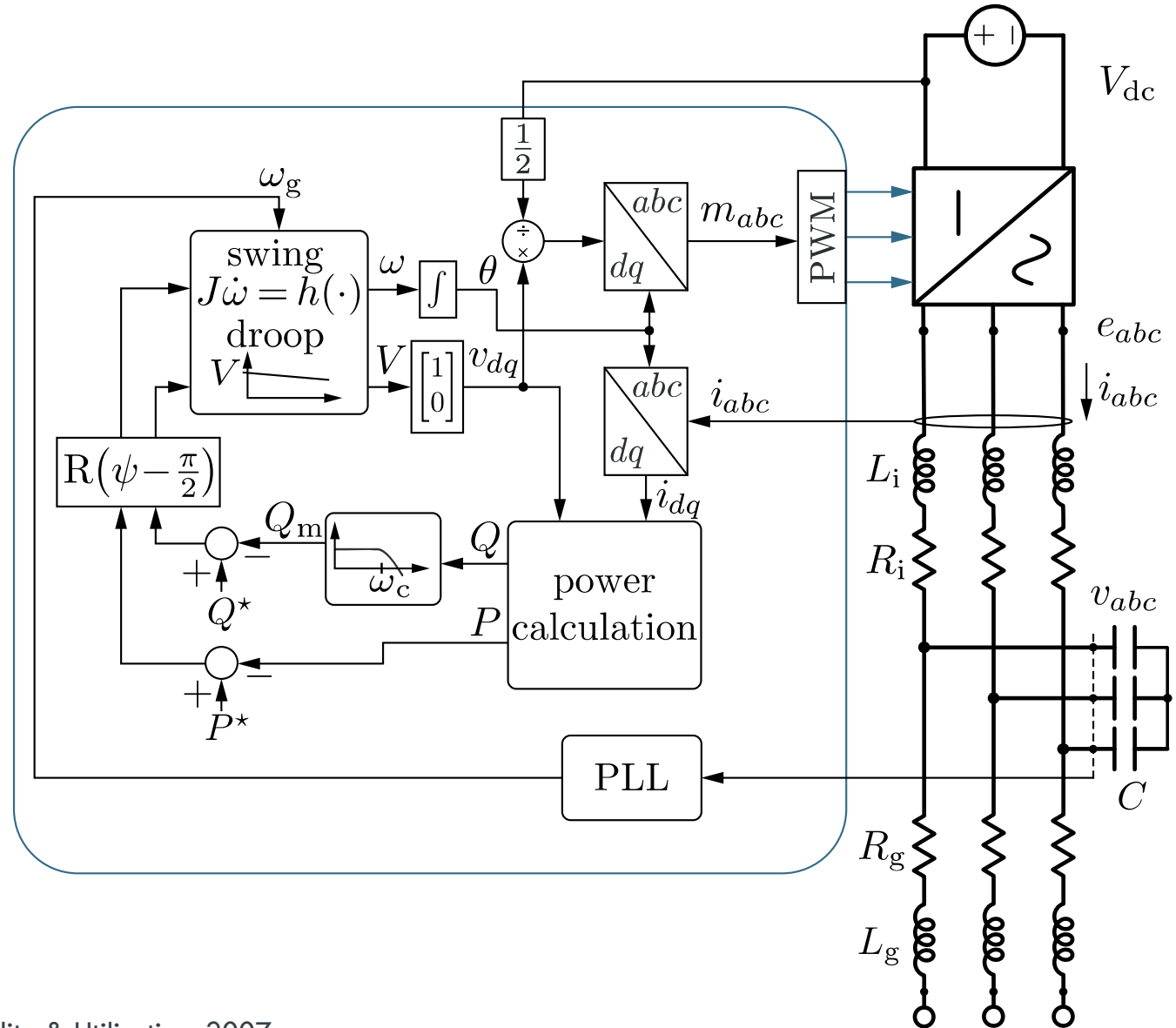
Control equations are given by

$$\begin{aligned}\omega &= \omega_0 + \frac{1}{d_{\text{f}}} \begin{bmatrix} 1 & 0 \end{bmatrix} \text{R}(\psi - \frac{\pi}{2}) \begin{bmatrix} P^\star - P_{\text{m}} \\ Q^\star - Q_{\text{m}} \end{bmatrix}, \\ V &= V_0 + \frac{1}{d_{\text{v}}} \begin{bmatrix} 0 & 1 \end{bmatrix} \text{R}(\psi - \frac{\pi}{2}) \begin{bmatrix} P^\star - P_{\text{m}} \\ Q^\star - Q_{\text{m}} \end{bmatrix}, \\ \frac{1}{\omega_{\text{c}}} \begin{bmatrix} \dot{P}_{\text{m}} \\ \dot{Q}_{\text{m}} \end{bmatrix} &= - \begin{bmatrix} P_{\text{m}} \\ Q_{\text{m}} \end{bmatrix} + \begin{bmatrix} P \\ Q \end{bmatrix}.\end{aligned}$$



## Approach #3: Virtual Synchronous Machine Control

- Emulate machine dynamics digitally
- Popular due to familiar behavior



[1] Beck, Hesse, "Virtual synchronous machine," Electrical Power Quality & Utilisation, 2007.

[2] D’Arco, Suul, Fosso, “A virtual synchronous machine implementation for distributed control of power converters in smartgrids,” EPSR, 2014.

# Approach #3: Virtual Synchronous Machine Control

Control equations are given by

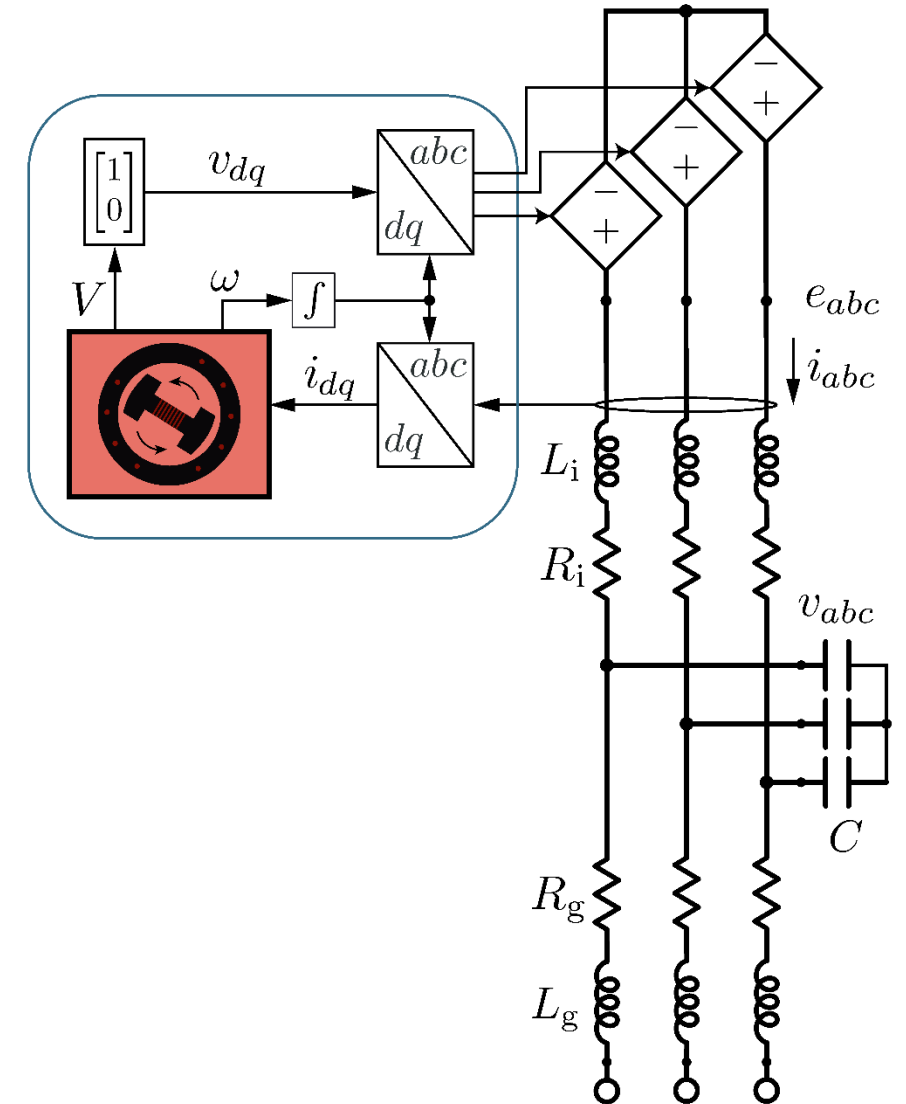
$$J\dot{\omega} = -\omega + \omega_0 + \frac{d_d}{d_f}(\omega_g - \omega) + \frac{1}{d_f} \begin{bmatrix} 1 & 0 \end{bmatrix} \mathbf{R}(\psi - \frac{\pi}{2}) \begin{bmatrix} P^* - P \\ Q^* - Q_m \end{bmatrix},$$

$$V = V_0 + \frac{1}{d_v} \begin{bmatrix} 0 & 1 \end{bmatrix} \mathbf{R}(\psi - \frac{\pi}{2}) \begin{bmatrix} P^* - P \\ Q^* - Q_m \end{bmatrix},$$

$$\frac{1}{\omega_0} \dot{\eta} = \begin{bmatrix} 0 & 1 \end{bmatrix} \mathbf{R}(\alpha) \mathbf{R}(\delta) \mathbf{T}(\omega_0 t) v,$$

$$\frac{1}{\omega_0} \dot{\alpha} = \frac{k_P}{\omega_0} \dot{\eta} + k_I \eta,$$

$$\frac{1}{\omega_c} \dot{Q}_m = -Q_m + Q.$$

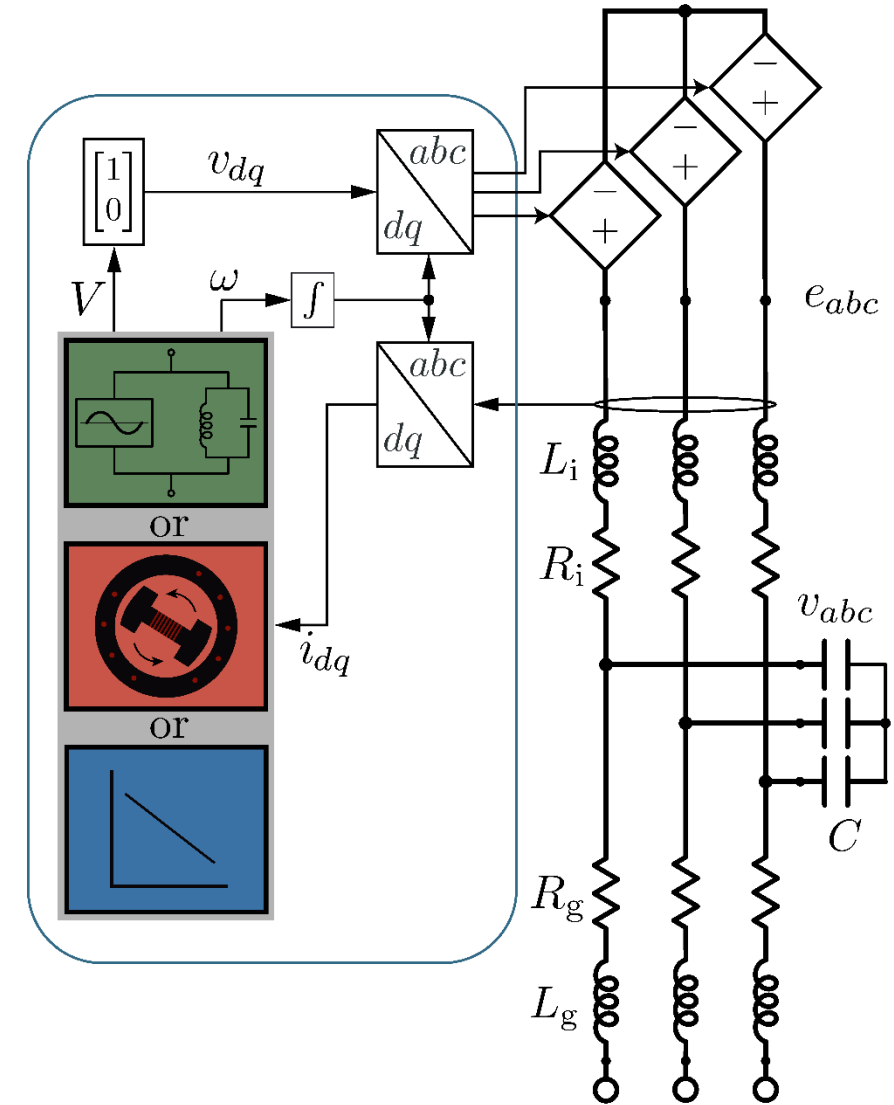




# A Universal & Unified GFM Model

## All 3 GFM types can be boiled down to

$$\begin{aligned}\tau_{\text{f}} \frac{\text{d}\omega}{\text{d}t} &= -\omega + \omega_0 + \kappa_{\text{d}} (\omega_{\text{g}} - \omega) \\ &\quad + \kappa_{\text{f}} \begin{bmatrix} 1 & 0 \end{bmatrix} \text{R}(\psi - \frac{\pi}{2}) \begin{bmatrix} P^{\star} - P_{\text{m}} \\ Q^{\star} - Q_{\text{m}} \end{bmatrix}, \\ \tau_{\text{v}} \frac{\text{d}V}{\text{d}t} &= f_{\text{v}}(V) + \kappa_{\text{v}} \begin{bmatrix} 0 & 1 \end{bmatrix} \text{R}(\psi - \frac{\pi}{2}) \begin{bmatrix} p^{\star} - p_{\text{m}} \\ q^{\star} - q_{\text{m}} \end{bmatrix}, \\ \frac{1}{\omega_0} \frac{\text{d}\eta}{\text{d}t} &= \begin{bmatrix} 0 & 1 \end{bmatrix} \text{R}(\alpha) \text{R}(\delta) \text{T}(\omega_0 t) v, \\ \frac{1}{\omega_0} \frac{\text{d}\alpha}{\text{d}t} &= \frac{k_{\text{P}}}{\omega_0} \dot{\eta} + k_{\text{I}} \eta, \\ \tau_{\text{P}} \begin{bmatrix} \dot{P}_{\text{m}} \\ \dot{Q}_{\text{m}} \end{bmatrix} &= - \begin{bmatrix} P_{\text{m}} \\ Q_{\text{m}} \end{bmatrix} + \begin{bmatrix} P \\ Q \end{bmatrix}.\end{aligned}$$



- [1] Johnson, et. al, "A generic primary-control model for grid-forming inverters: Towards interoperable operation & control," HICSS, 2021.
- [2] Ajala, Baeckeland, Dhople, Dominguez-Garcia, "Uncovering the Kuramoto model from full-order models of grid-forming inverter-based power networks," CDC, 2021.

# A Universal & Unified GFM Model

All 3 GFM types can be boiled down to

$$\begin{aligned}\tau_f \frac{d\omega}{dt} &= -\omega + \omega_0 + \kappa_d (\omega_g - \omega) \\ &\quad + \kappa_f \begin{bmatrix} 1 & 0 \end{bmatrix} \mathbf{R}(\psi - \frac{\pi}{2}) \begin{bmatrix} P^* - P_m \\ Q^* - Q_m \end{bmatrix}, \\ \tau_v \frac{dV}{dt} &= f_v(V) + \kappa_v \begin{bmatrix} 0 & 1 \end{bmatrix} \mathbf{R}(\psi - \frac{\pi}{2}) \begin{bmatrix} p^* - p_m \\ q^* - q_m \end{bmatrix}, \\ \frac{1}{\omega_0} \frac{d\eta}{dt} &= \begin{bmatrix} 0 & 1 \end{bmatrix} \mathbf{R}(\alpha) \mathbf{R}(\delta) \mathbf{T}(\omega_0 t) v, \\ \frac{1}{\omega_0} \frac{d\alpha}{dt} &= \frac{k_P}{\omega_0} \dot{\eta} + k_I \eta, \\ \tau_P \begin{bmatrix} \dot{P}_m \\ \dot{Q}_m \end{bmatrix} &= - \begin{bmatrix} P_m \\ Q_m \end{bmatrix} + \begin{bmatrix} P \\ Q \end{bmatrix}.\end{aligned}$$

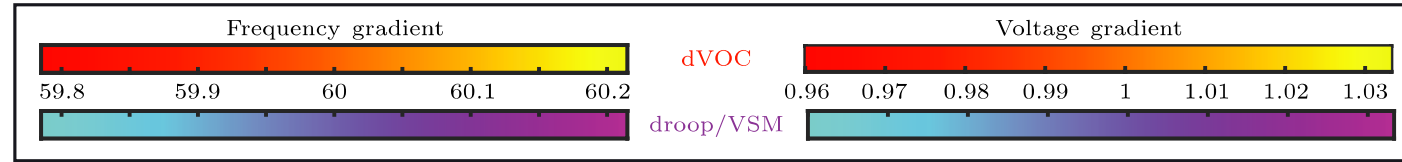
where the parameters are

	$\tau_f$	$\tau_v$	$\tau_P$	$\kappa_d$	$\kappa_f$	$\kappa_v$	$f_v(e^*)$
<i>droop</i>	0	0	$\frac{1}{\omega_c}$	0	$\frac{1}{d_f}$	$\frac{1}{d_v}$	$-V + V_0$
<i>VSM</i>	$J$	0	$\frac{1}{\omega_c}$	$\frac{d_d}{d_f}$	$\frac{1}{d_f}$	$\frac{1}{d_v}$	$-V + V_0$
<i>dVOC</i>	0	$\frac{1}{\omega_0}$	0	0	$\frac{\omega_0 \kappa_1}{V^2}$	$\frac{\kappa_1}{V}$	$\kappa_2(-V^2 + V_0^2)V$

[1] Johnson, et. al, “A generic primary-control model for grid-forming inverters: Towards interoperable operation & control,” HICSS, 2021.

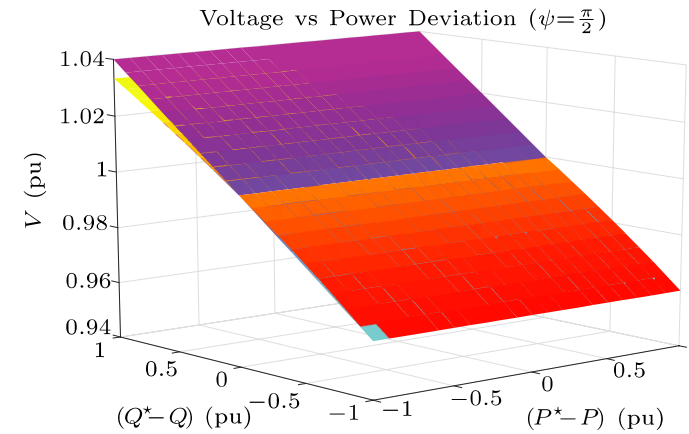
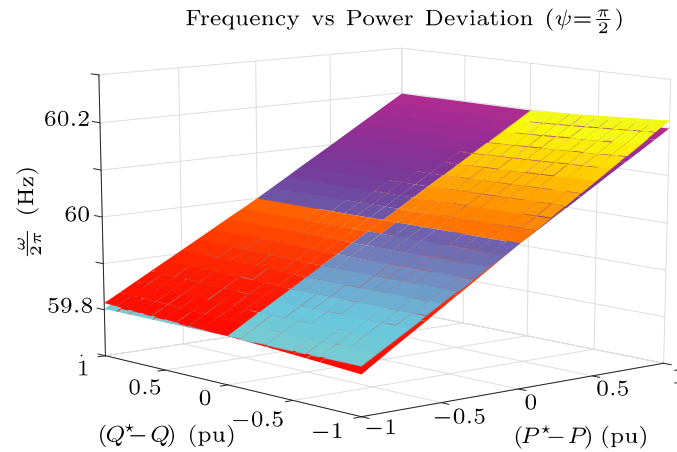
[2] Ajala, Baeckeland, Dhople, Dominguez-Garcia, “Uncovering the Kuramoto model from full-order models of grid-forming inverter-based power networks,” CDC, 2021.

# Steady-state Performance of GFM Types



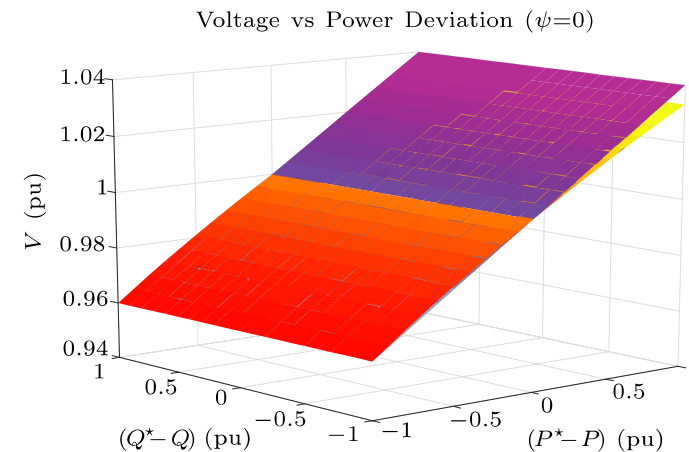
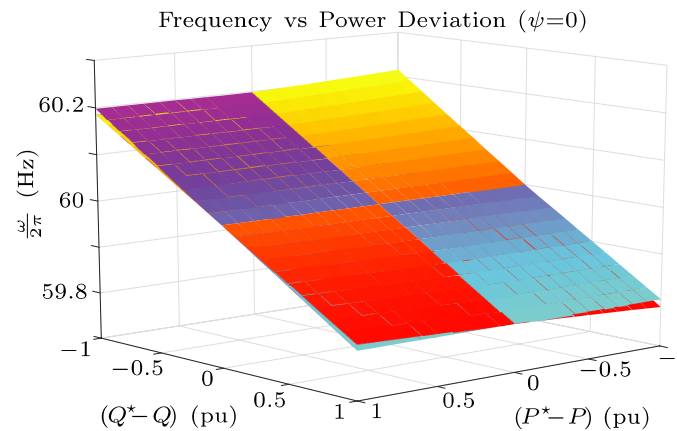
inductive lines

$$\psi = \frac{\pi}{2}$$



resistive lines

$$\psi = 0$$



[1] Johnson, et. al, "A generic primary-control model for grid-forming inverters: Towards interoperable operation & control," HICSS, 2021.

[2] Ajala, Baeckeland, Dhople, Dominguez-Garcia, "Uncovering the Kuramoto model from full-order models of grid-forming inverter-based power networks," CDC, 2021.

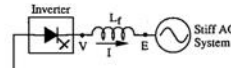
# Unifying 25+ Years of GFM Technologies

1993

## Control of Parallel Connected Inverters in Standalone ac Supply Systems

Mukul C. Chandorkar, *Student Member, IEEE*, Deepakraj M. Divan, *Member, IEEE*, and Rambabu Adapa, *Senior Member, IEEE*

**Abstract**—A scheme for controlling parallel-connected inverters in a standalone ac supply system is presented in this paper. This scheme is suitable for control of inverters in distributed source environments such as in isolated ac systems, large and



2014

## Synchronization of Parallel Single-Phase Inverters With Virtual Oscillator Control

Brian B. Johnson, *Member, IEEE*, Sairaj V. Dhople, *Member, IEEE*, Abdullah O. Hamadeh, and Philip T. Krein, *Fellow, IEEE*

**Abstract**—A method to synchronize and control a system of parallel single-phase inverters without communication is presented. Inspired by the phenomenon of synchronization in networks of coupled oscillators, we propose that each inverter be controlled

proportion to their ratings. Focused on these challenges, this paper presents a method to synchronize and control a system of parallel single-phase inverters without communication. It is important to clarify that synchronization in our setting corre

2007

## Virtual Synchronous Machine

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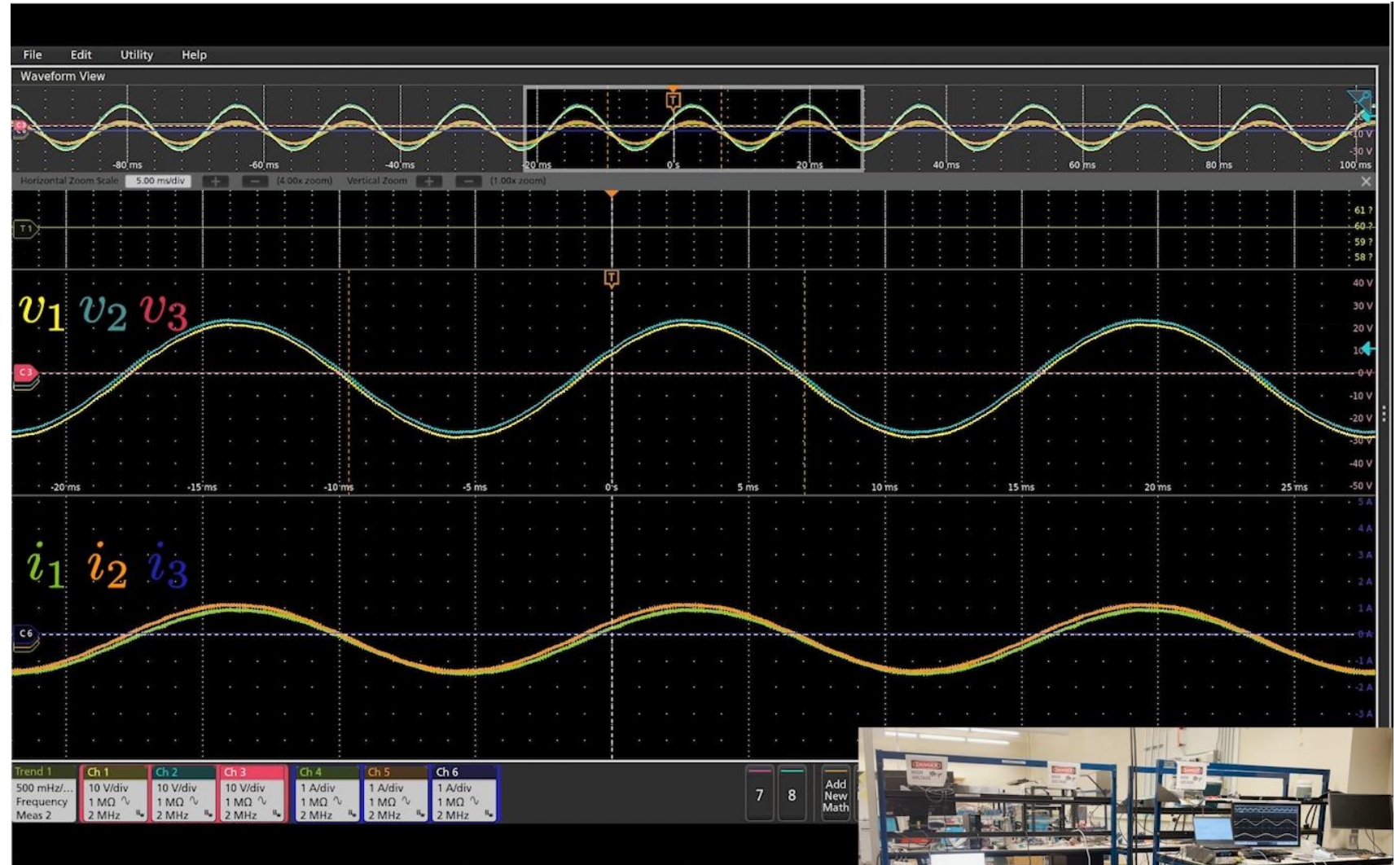
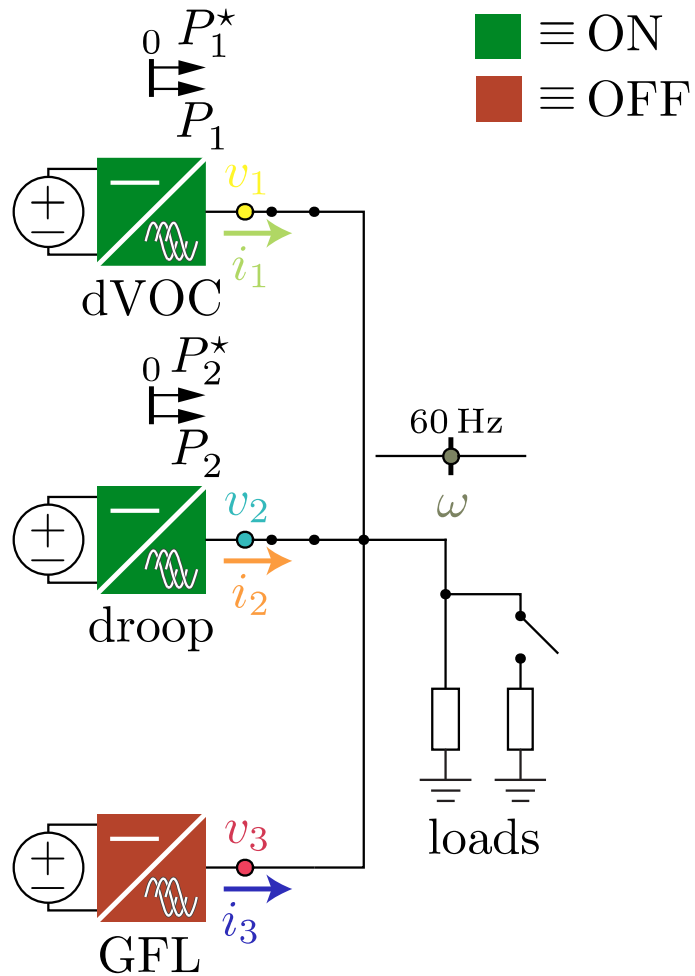
Demands in the area of electrical energy generation and distribution, as a result of energy policies, are leading to far reaching changes in the structure of the energy supply, which is

cause or promote the appearance and mitigation of various disturbances.

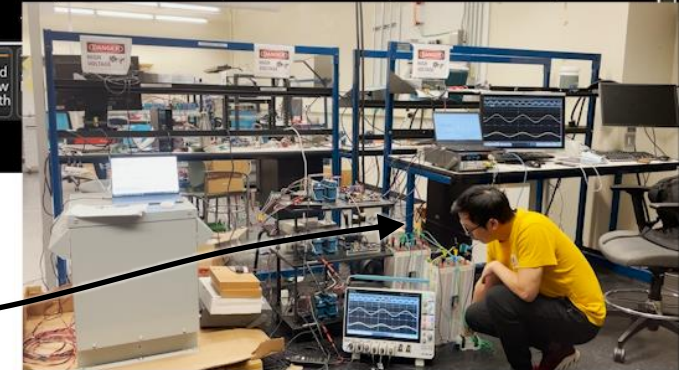
$$\begin{aligned}\tau_f \frac{d\omega}{dt} &= -\omega + \omega_0 + \kappa_d (\omega_g - \omega) \\ &\quad + \kappa_f \begin{bmatrix} 1 & 0 \end{bmatrix} R(\psi - \frac{\pi}{2}) \begin{bmatrix} P^* - P_m \\ Q^* - Q_m \end{bmatrix}, \\ \tau_v \frac{dV}{dt} &= f_v(V) + \kappa_v \begin{bmatrix} 0 & 1 \end{bmatrix} R(\psi - \frac{\pi}{2}) \begin{bmatrix} p^* - p_m \\ q^* - q_m \end{bmatrix}, \\ \frac{1}{\omega_0} \frac{d\eta}{dt} &= \begin{bmatrix} 0 & 1 \end{bmatrix} R(\alpha) R(\delta) T(\omega_0 t) v, \\ \frac{1}{\omega_0} \frac{d\alpha}{dt} &= \frac{k_P}{\omega_0} \dot{\eta} + k_I \eta, \\ \tau_P \begin{bmatrix} \dot{P}_m \\ \dot{Q}_m \end{bmatrix} &= - \begin{bmatrix} P_m \\ Q_m \end{bmatrix} + \begin{bmatrix} P \\ Q \end{bmatrix},\end{aligned}$$

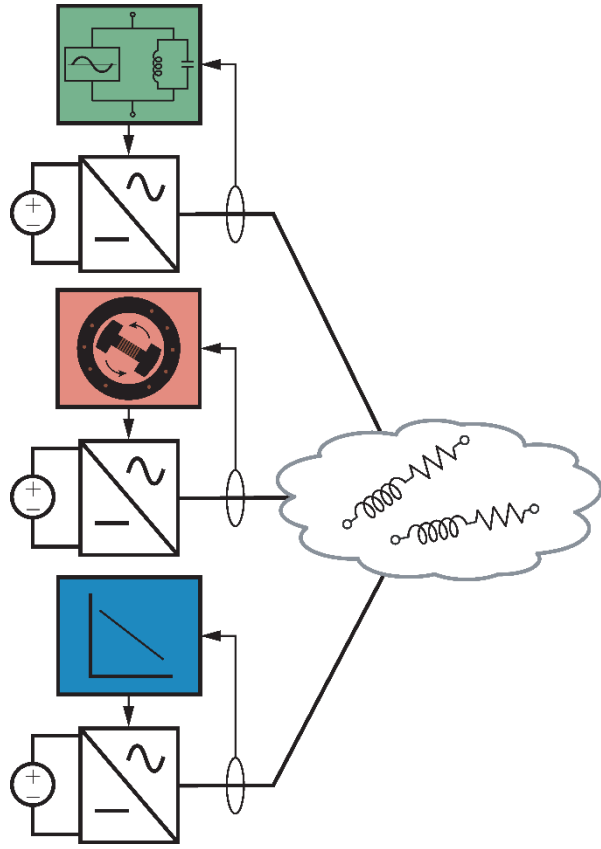
A unified model

# Showing Interoperability with a Mix of Control Types

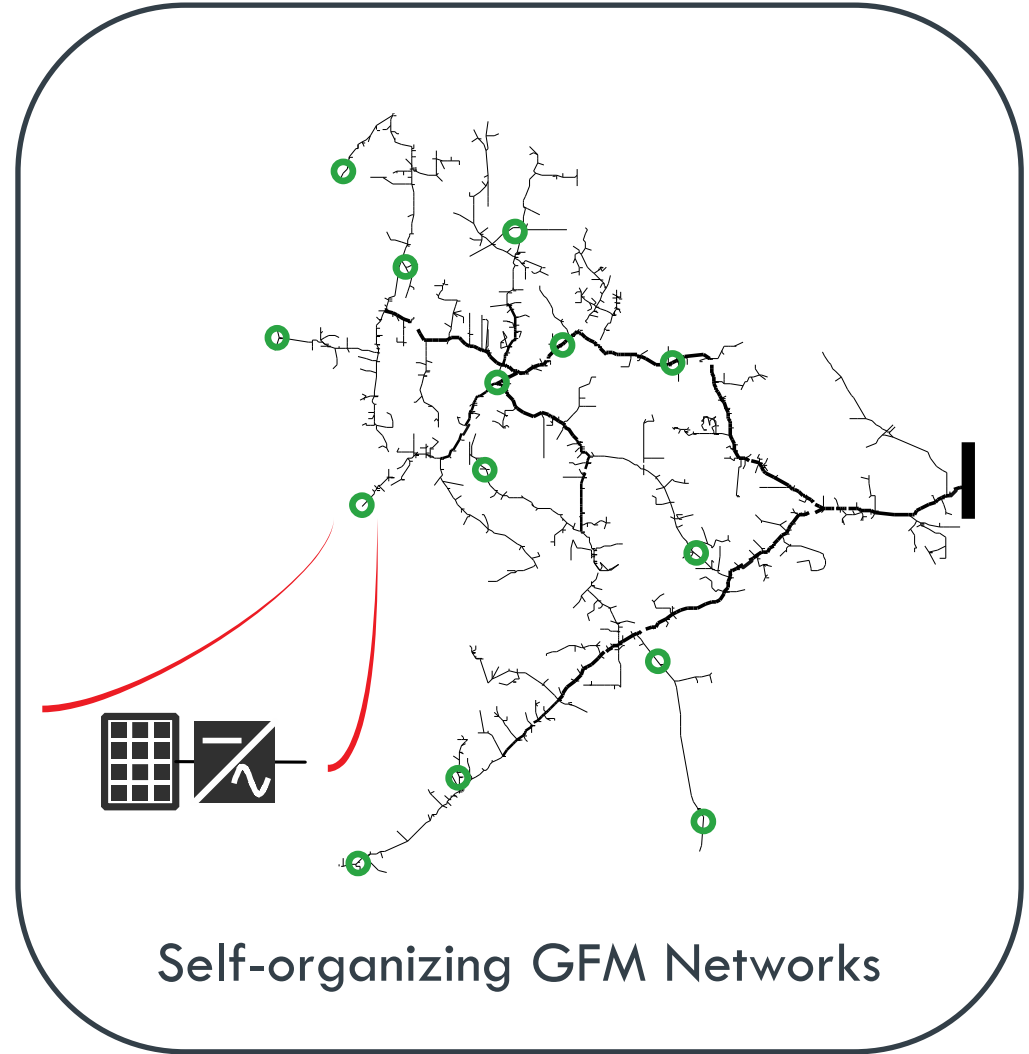


Weiqian Cai





Interoperability



Self-organizing GFM Networks



# A Single-phase GFM Commercial Product

## Features of the inverter building block

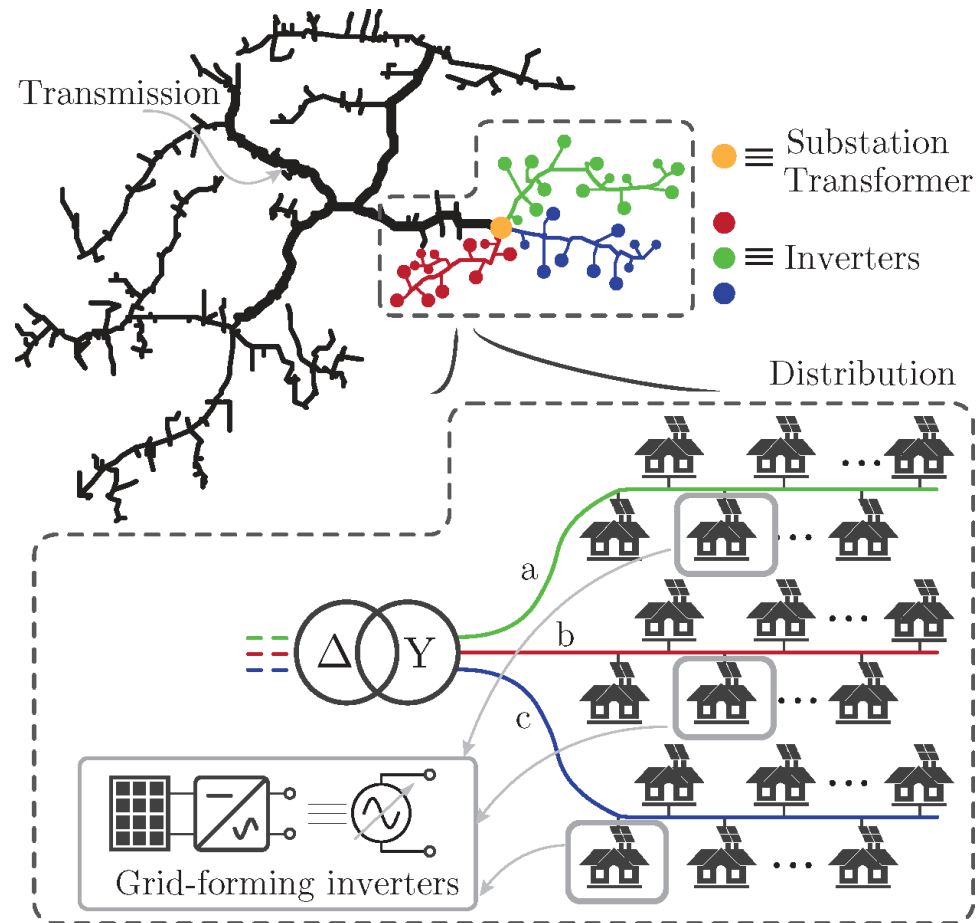
- 300 VA single-phase with droop-based GFM controls
- Bidirectional converter can interface PV or batteries



Over 39M+ Enphase inverters shipped for 12 GW of capacity as of September 2021

# Building Three-phase Systems with Single-phase GFM Units

Can swarms of decentralized single-phase GFM units self-organize into a three-phase system?



Could this system maintain phase balancing when islanded?

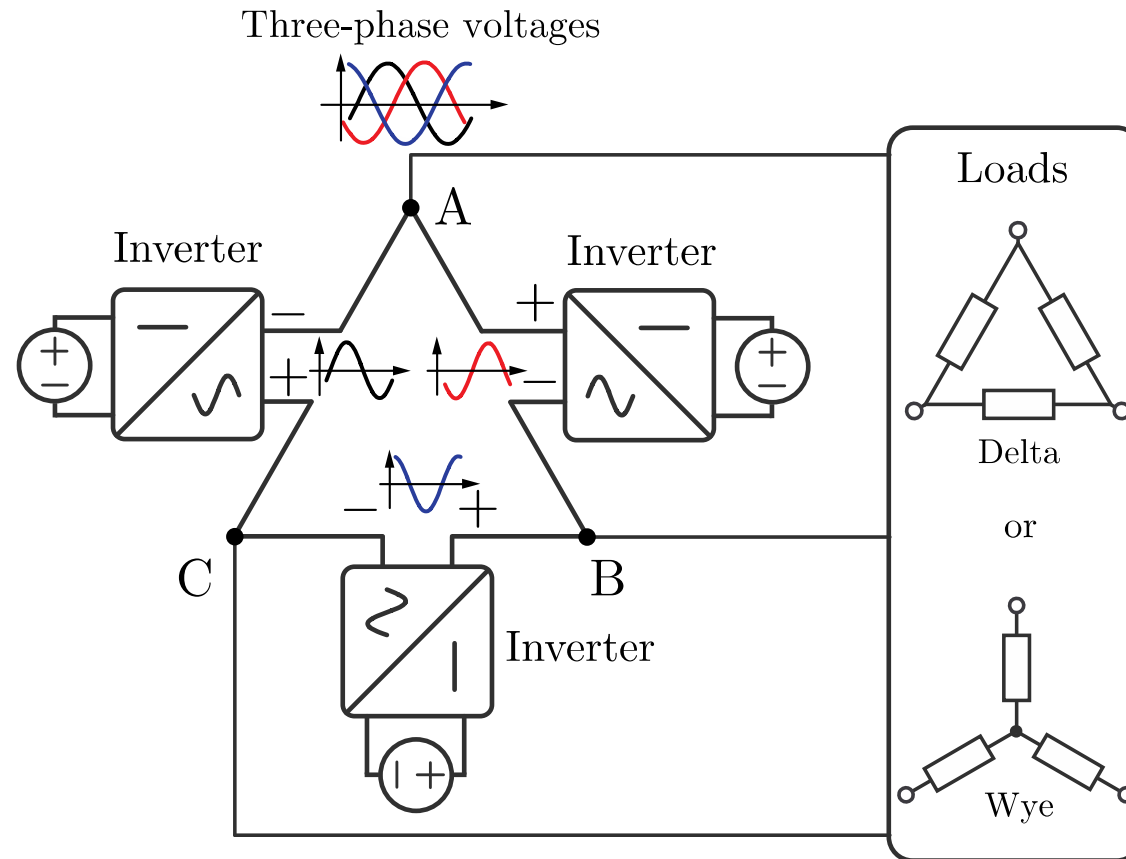


# From Industry Challenges to Foundational Research

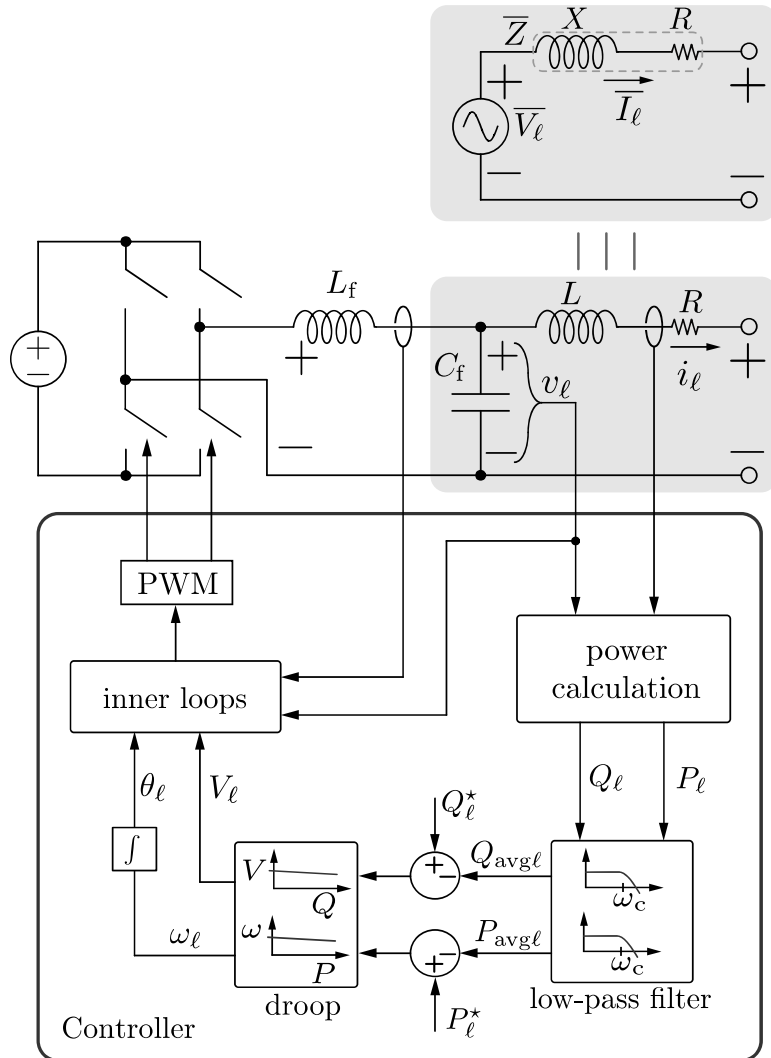
Industry partners have already prioritized three-phase system operation with single-phase units



Donny  
Zimmanck



# Single-phase Inverter Model



A single-phase droop-controlled inverter

The  $\ell \in 1, 2, 3$  inverter has the droop laws

$$V_\ell = V_{\text{nom}} - m_q (Q_{\text{avg},\ell} - Q_\ell^*),$$

$$\omega_\ell = \omega_{\text{nom}} - m_p (P_{\text{avg},\ell} - P_\ell^*),$$

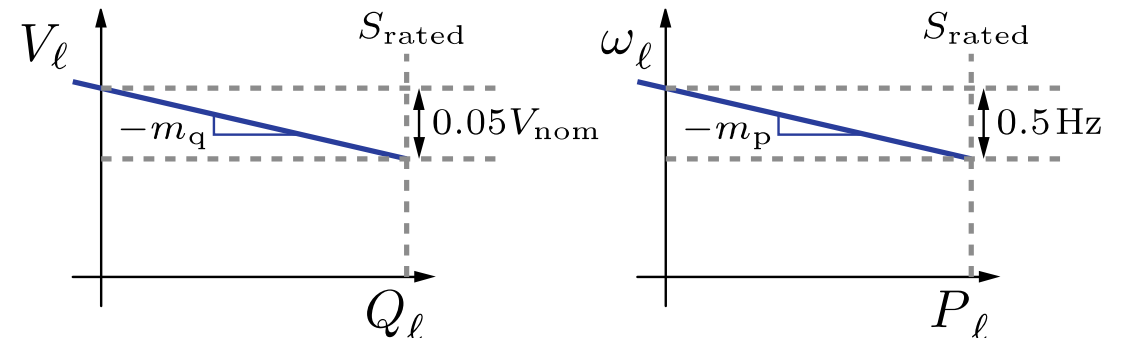
where

$$\dot{P}_{\text{avg},\ell} = \omega_c (P_\ell - P_{\text{avg},\ell}),$$

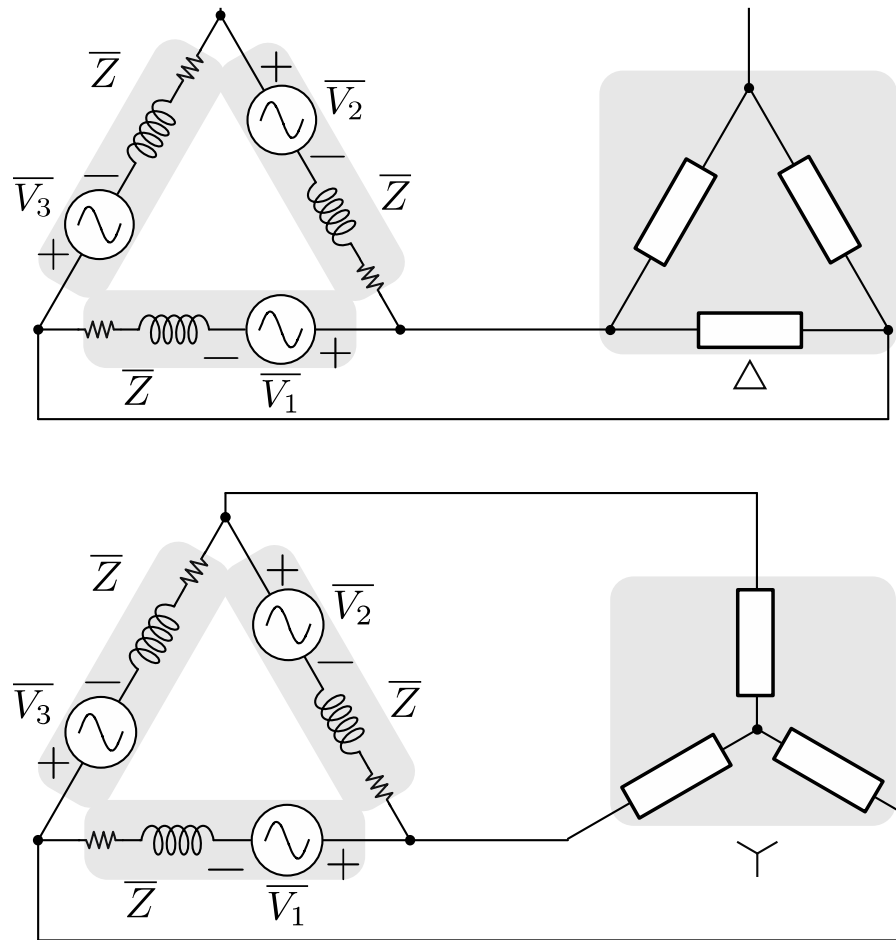
$$\dot{Q}_{\text{avg},\ell} = \omega_c (Q_\ell - Q_{\text{avg},\ell}),$$

and slopes  $m_q = 0.05 \frac{V_{\text{nom}}}{S_{\text{rated}}}$ ,  $m_p = \frac{2\pi \times 0.5}{S_{\text{rated}}}$ . This gives

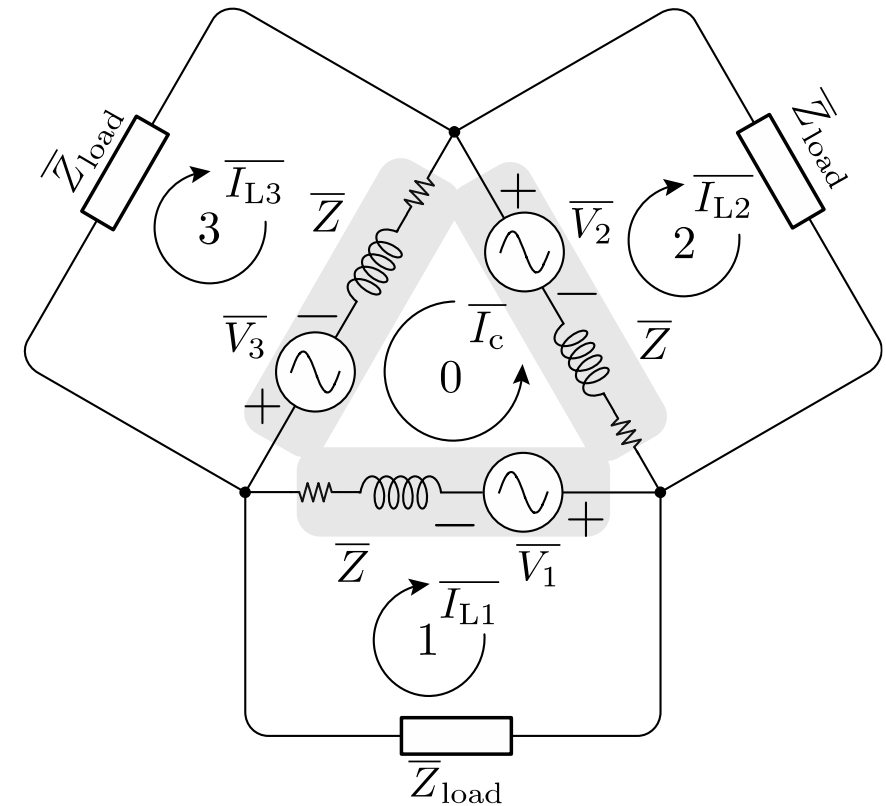
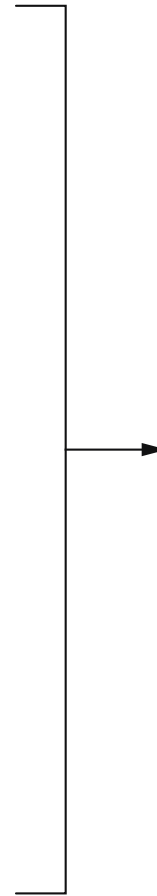
the following in steady-state:



# A Common Framework For Various Load Configurations



inverters with  $\Delta$  &  $\Upsilon$  loads



circuit model with mesh loops

# A System Circuit Model

Circuit laws + control dynamics give

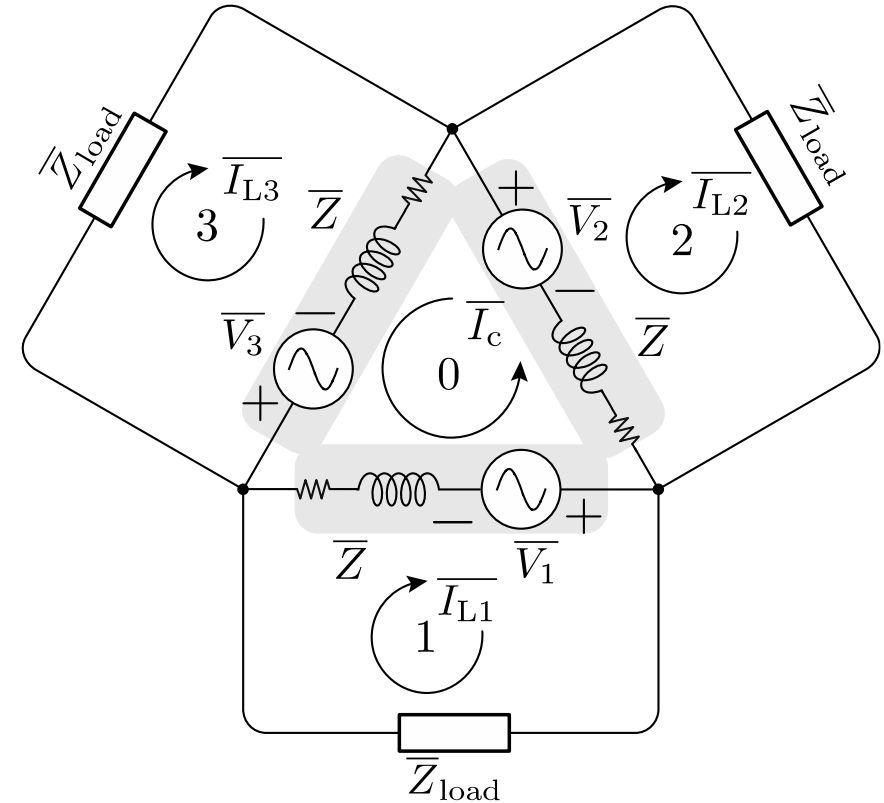
$$V_\ell = V_{\text{nom}} - m_q \sum_{k=1}^3 \frac{V_k V_\ell}{|\bar{Z}_{\text{loop}}|} \sin(\theta_{k\ell} - \phi),$$

$$\dot{\theta}_\ell = \omega_{\text{nom}} - m_p \sum_{k=1}^3 \frac{V_k V_\ell}{|\bar{Z}_{\text{loop}}|} \cos(\theta_{k\ell} - \phi).$$

where  $\bar{Z}_{\text{loop}} = 3\bar{Z}$  and  $\phi = \tan^{-1}(X/R)$ . Next,  
focus on angle difference dynamics

$$\dot{\theta}_{21} = \dot{\theta}_2 - \dot{\theta}_1,$$

$$\dot{\theta}_{31} = \dot{\theta}_3 - \dot{\theta}_1.$$

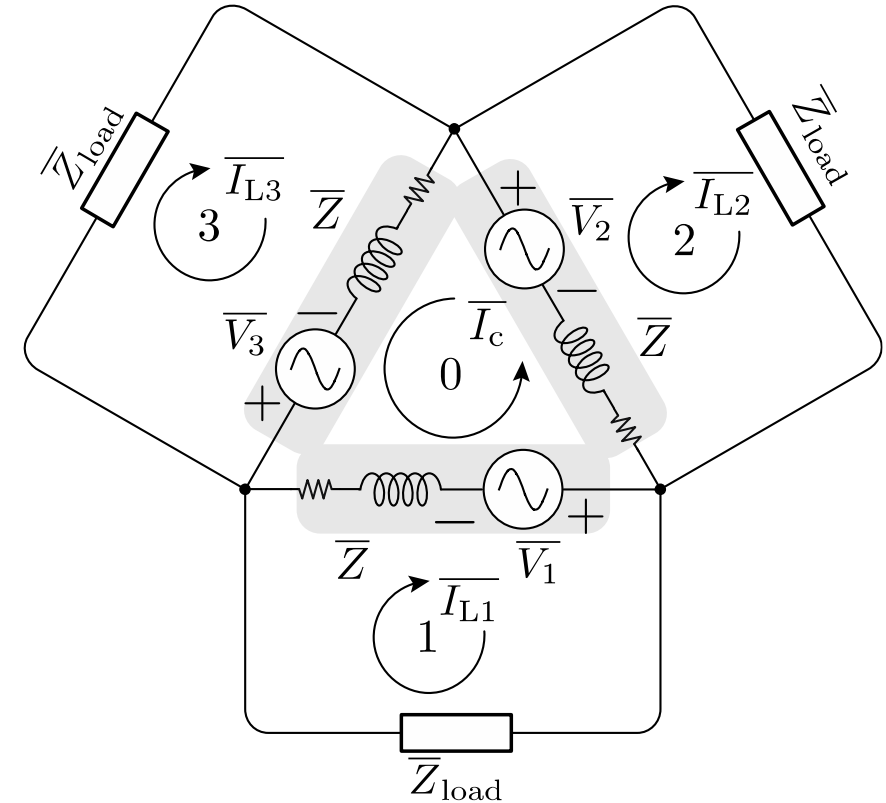


circuit model with mesh loops

# Nonlinear Angular Dynamics

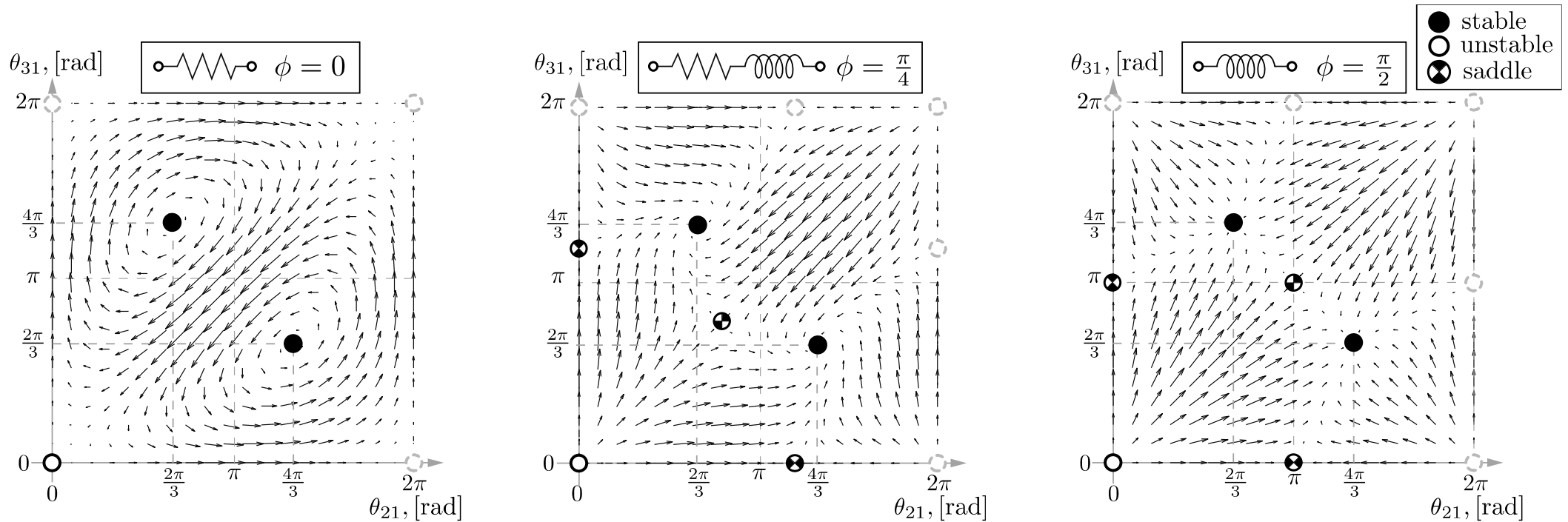
Algebra gives us

$$\begin{aligned}\dot{\theta}_{21} = & \frac{m_p V_2 V_1}{|\bar{Z}_{\text{loop}}|} \cos(\theta_{21} - \phi) + \frac{m_p V_3 V_1}{|\bar{Z}_{\text{loop}}|} \cos(\theta_{31} - \phi) \\ & - \frac{m_p V_2 V_3}{|\bar{Z}_{\text{loop}}|} \cos(\theta_{32} - \phi) - \frac{m_p V_2 V_1}{|\bar{Z}_{\text{loop}}|} \cos(\theta_{12} - \phi), \\ \dot{\theta}_{31} = & \frac{m_p V_3 V_1}{|\bar{Z}_{\text{loop}}|} \cos(\theta_{31} - \phi) + \frac{m_p V_2 V_1}{|\bar{Z}_{\text{loop}}|} \cos(\theta_{21} - \phi) \\ & + \frac{m_p V_2 V_3}{|\bar{Z}_{\text{loop}}|} \cos(\theta_{23} - \phi) + \frac{m_p V_1 V_3}{|\bar{Z}_{\text{loop}}|} \cos(\theta_{13} - \phi).\end{aligned}$$



circuit model with mesh loops

# Angle Difference Dynamics

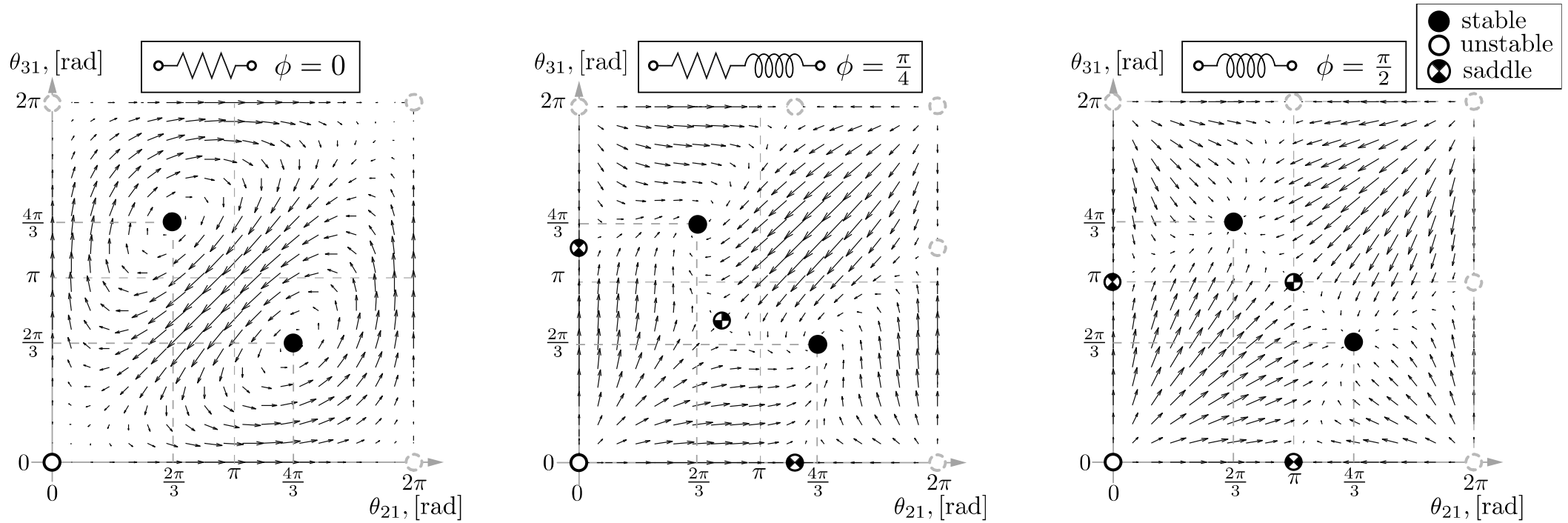


Substitute voltage droop laws, ignore  $m_p^2, m_q^2, m_p m_q$  terms, and let  $K = m_p V_{\text{nom}}^2 / |\bar{Z}_{\text{loop}}|$  to get:

$$\dot{\theta}_{21} \approx K(2 \sin \theta_{21} + \sin \theta_{31} + \sin (\theta_{21} - \theta_{31})) \sin \phi + K(\cos \theta_{31} - \cos (\theta_{21} - \theta_{31})) \cos \phi,$$

$$\dot{\theta}_{31} \approx K(2 \sin \theta_{31} + \sin \theta_{21} + \sin (\theta_{31} - \theta_{21})) \sin \phi + K(\cos \theta_{21} - \cos (\theta_{31} - \theta_{21})) \cos \phi.$$

# Angle Difference Dynamics



The equilibria are below, where  $\sigma_1 = 2\pi + 2 \tan^{-1}(-3 \tan \phi)$ ,  $\sigma_2 = 2 \tan^{-1}(3 \tan \phi)$ .

$$(\theta_{21,\text{eq}}, \theta_{31,\text{eq}})^{\bullet} = \left\{ \left( \frac{2\pi}{3}, \frac{4\pi}{3} \right), \left( \frac{4\pi}{3}, \frac{2\pi}{3} \right) \right\},$$

$$(\theta_{21,\text{eq}}, \theta_{31,\text{eq}})^{\circ} = \{(0, 0)\},$$

$$(\theta_{21,\text{eq}}, \theta_{31,\text{eq}})^{\otimes} = \{(0, \sigma_1), (\sigma_1, 0), (\sigma_2, \sigma_2)\},$$

# Small-signal Modeling

The linearized system is

$$\begin{bmatrix} \Delta \dot{\theta}_{21} \\ \Delta \dot{\theta}_{31} \end{bmatrix} = \mathcal{J}(\theta_{21,\text{eq}}, \theta_{31,\text{eq}}) \begin{bmatrix} \Delta \theta_{21} \\ \Delta \theta_{31} \end{bmatrix},$$

where

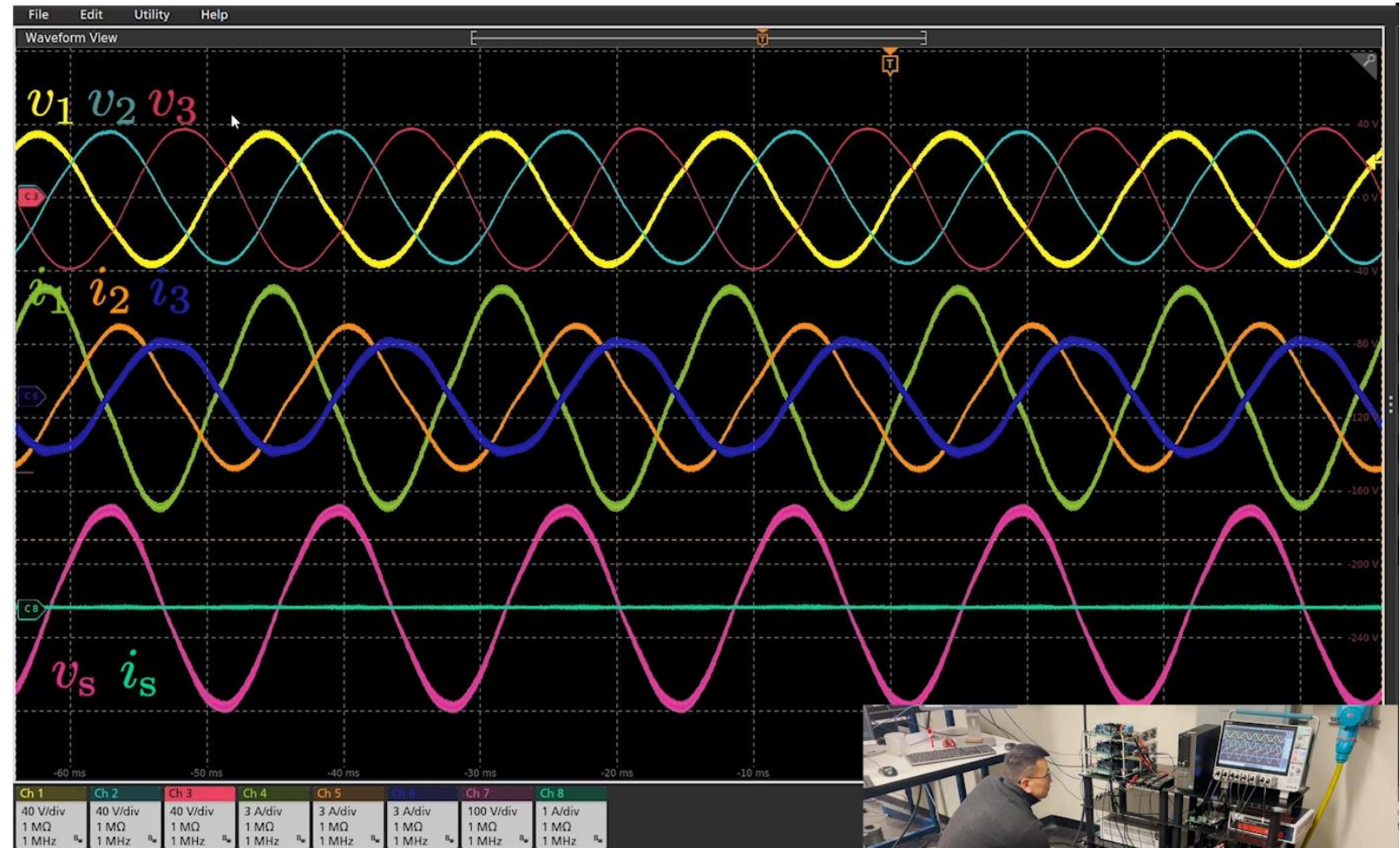
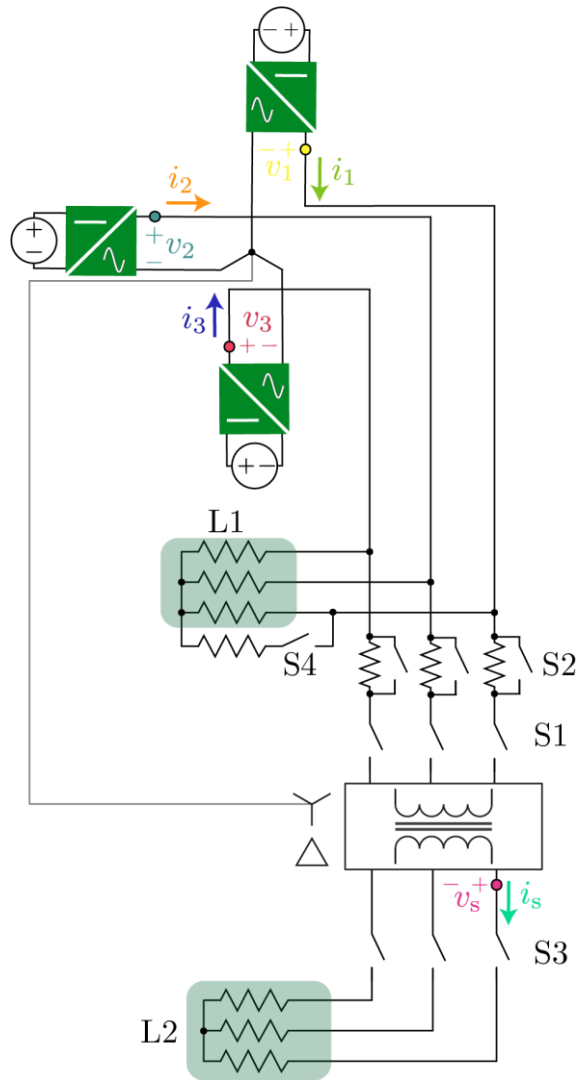
$$\mathcal{J}(\theta_{21,\text{eq}}, \theta_{31,\text{eq}}) = K \begin{bmatrix} (2 \cos \theta_{21,\text{eq}} + \cos (\theta_{21,\text{eq}} - \theta_{31,\text{eq}})) \sin \phi & | & (\cos \theta_{31,\text{eq}} - \cos (\theta_{21,\text{eq}} - \theta_{31,\text{eq}})) \sin \phi \\ + \sin (\theta_{21,\text{eq}} - \theta_{31,\text{eq}}) \cos \phi & & + (\sin (\theta_{31,\text{eq}} - \theta_{21,\text{eq}}) - \sin \theta_{31,\text{eq}}) \cos \phi \\ \hline (\cos \theta_{21,\text{eq}} - \cos (\theta_{31,\text{eq}} - \theta_{21,\text{eq}})) \sin \phi & | & (2 \cos \theta_{31,\text{eq}} + \cos (\theta_{31,\text{eq}} - \theta_{21,\text{eq}})) \sin \phi \\ + (\sin (\theta_{21,\text{eq}} - \theta_{31,\text{eq}}) - \sin \theta_{21,\text{eq}}) \cos \phi & & + \sin (\theta_{31,\text{eq}} - \theta_{21,\text{eq}}) \cos \phi \end{bmatrix},$$

and the eigenvalues for all equilibria within  $\phi \in (0, \pi/2]$  are

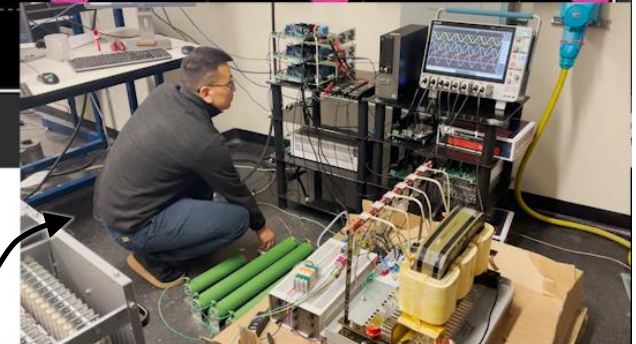
$$\begin{aligned} \lambda_{1,2}^{\bullet} &= -\frac{3}{2}K(\sin \phi \pm j \cos \phi), \longrightarrow \Re(\lambda_{1,2}^{\bullet}) < 0 \longrightarrow \text{stable,} \\ \lambda_{1,2}^{\circ} &= 3K \sin \phi, \longrightarrow \Re(\lambda_{1,2}^{\circ}) > 0 \longrightarrow \text{unstable,} \\ \lambda_1^{\ominus} &= -3K \sin \phi, \quad \lambda_2^{\ominus} = 9K \frac{1 + \tan^2 \phi}{1 + 9 \tan^2 \phi} \sin \phi, \longrightarrow \Re(\lambda_1^{\ominus}) < 0 \text{ and } \Re(\lambda_2^{\ominus}) > 0 \longrightarrow \text{saddle.} \end{aligned}$$



# Self-balancing Single-phase GFM Hardware Results

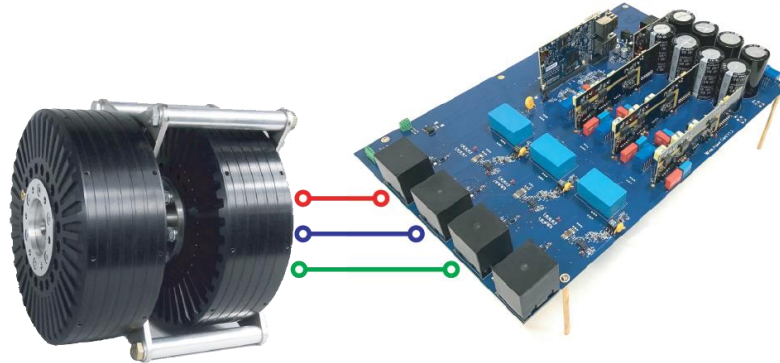


Dr. Minghui Lu

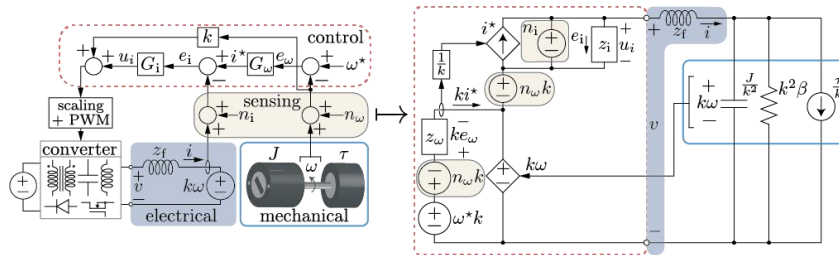


# Thanks for Your Attention!

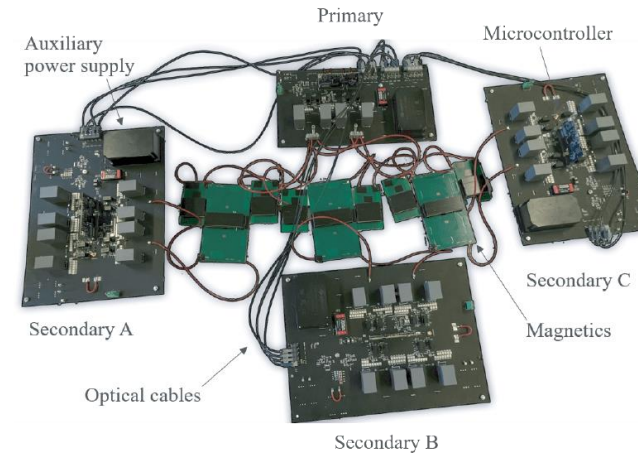
My other work building & analyzing future energy systems at all scales



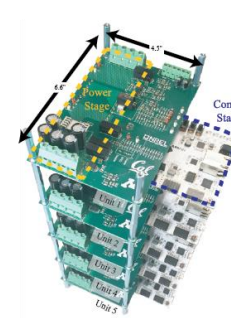
electromechanics



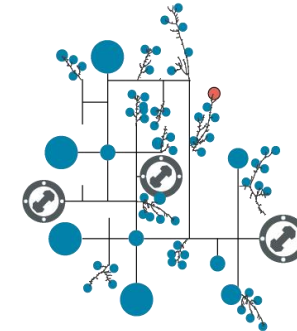
multiphysics modeling



medium-voltage electronics



dc-dc converters



system dynamics